

## • SOLIDS & SEMICONDUCTOR DEVICES •

Electronic instruments are being utilized in various fields like telecommunication, entertainment, computers, nuclear physics and many more. Although the history started with the advent of vacuum tubes, however the rapid advancement in electronics which we see today is due to the valuable contributions of semiconductor devices.

Semiconductor devices are not only small in size, consumes less power, have long life times and are more efficient than vacuum tubes but also are of low cost. That is why these have replaced vacuum tubes nearly in all applications. As an Ex. we can consider the case of a computer. In early days, the vacuum tube based computers were as big as the size of a room and were capable of performing simple calculations only. At present the personal computer (PC) that you see in laboratory or at your home is much smaller in size and capable of performing many operations. Needless to say this is possible because of the advances in semiconductor technology.

We will learn the basic concept of semiconductors. This will enable us to understand the operation of many semiconductor devices and then we will be discussing few semiconductor devices like diode, transistor along with their applications.

### CLASSIFICATION OF METALS, CONDUCTORS AND SEMICONDUCTORS

#### On the basis of conductivity

On the basis of the relative values of electrical conductivity ( $\sigma$ ) or resistivity ( $\rho = 1/\sigma$ ), the solids are broadly classified as:

**(i) Metals:** They possess very low resistivity (or high conductivity).

$$\rho \sim 10^{-2} - 10^{-8} \Omega \text{ m}$$

$$\sigma \sim 10^2 - 10^8 \text{ S m}^{-1}$$

**(ii) Semiconductors:** They have resistivity or conductivity intermediate to metals and insulators.

$$\rho \sim 10^{-5} - 10^6 \Omega \text{ m}$$

$$\sigma \sim 10^5 - 10^{-6} \text{ S m}^{-1}$$

**(iii) Insulators:** They have high resistivity (or low conductivity).

$$\rho \sim 10^{11} - 10^{19} \Omega \text{ m}$$

$$\sigma \sim 10^{-11} - 10^{-19} \text{ S m}^{-1}$$

The values of  $\rho$  and  $\sigma$  given above are indicative of magnitude and could well go outside the ranges as well. Relative values of the resistivity are not the only criteria for distinguishing metals, insulators and semiconductors from each other. There are some other differences, which will become clear as we go along in this chapter. Our interest in this chapter is in the study of semiconductors which could be:

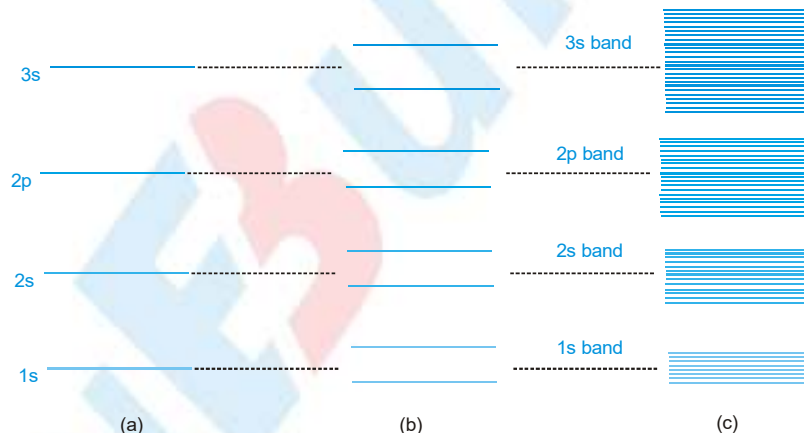
- (i) **Elemental semiconductors:** Si and Ge
- (ii) **Compound semiconductors:** Ex.s are:
  - ❖ Inorganic: CdS, GaAs, CdSe, InP, etc.
  - ❖ Organic: anthracene, doped phthalocyanines, etc.
  - ❖ Organic polymers: polypyrrole, polyaniline, polythiophene, etc.

Most of the currently available semiconductor devices are based on elemental semiconductors Si or Ge and compound inorganic semiconductors. However, after 1990, a few semiconductor devices using organic semiconductors and semiconducting polymers have been developed signalling the birth of a futuristic technology of polymerelectronics and molecular-electronics. In this chapter, we will restrict ourselves to the study of inorganic semiconductors, particularly elemental semiconductors Si and Ge. The general concepts introduced here for discussing the elemental semiconductors, by-and-large, apply to most of the compound semiconductors as well.

## 1. Energy Levels and Energy Bands in Solids

The electrons of an isolated atom are restricted to well defined energy levels. The maximum number of electrons which can be accommodated in any level is determined by the Pauli exclusion principle. The electrons belonging to the outermost energy level are called valence electrons. For Ex., the electronic configuration of sodium (atomic number 11) is  $1s^2 2s^2 2p^6 3s^1$ , here the electron belonging to the 3s level is the valence electron. Most of the solids including metals with which we are familiar occur in crystalline form. As we know a crystal is a regular periodic arrangement of atoms separated from each other by very small distance called lattice constant. The value of lattice constant is different for different crystalline solids, however it is of the order of linear dimension of atoms ( $\sim \text{\AA}$ ). Obviously at such a short separation between various neighbouring atoms, electrons in an atom cannot only be subjected to the Coulombic force of the nucleus of this atom but also obey Coulombic forces due to nuclei and electrons of the neighbouring atoms. In fact it is this interaction which results in the bonding between various atoms which leads to the formation of crystals.

When atoms are interacting (such as in crystal) then the energy level scheme for the individual atoms as shown in figure(a) does not quite hold. The interaction between atoms affects the electron energy levels, as a result there occurs a splitting of energy levels belonging to various atoms. To understand this phenomenon in more clear terms, let us first consider the simplest case of two interacting identical atoms. Let us assume that initially they are far apart i.e. the forces of interaction between them can be neglected. [If the distance between two atoms is much larger ( $\sim 50 \text{\AA}$ ) compared to their linear dimensions ( $\sim 10 \text{\AA}$ ) this assumption is reasonably correct]. In such a case we may treat them as isolated with energy levels like that for the case of an isolated atom as shown in figure(a).

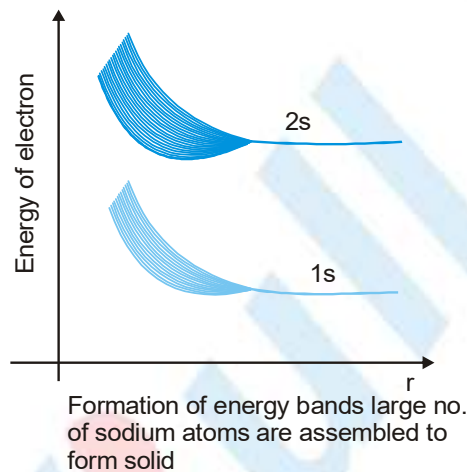
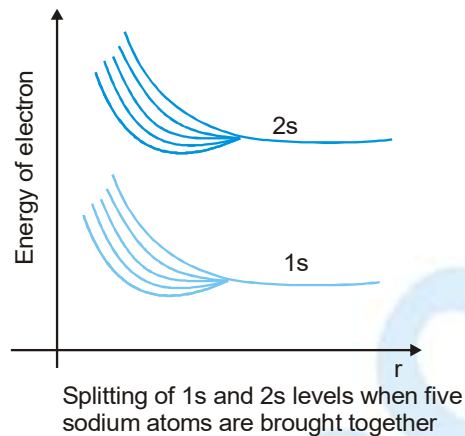
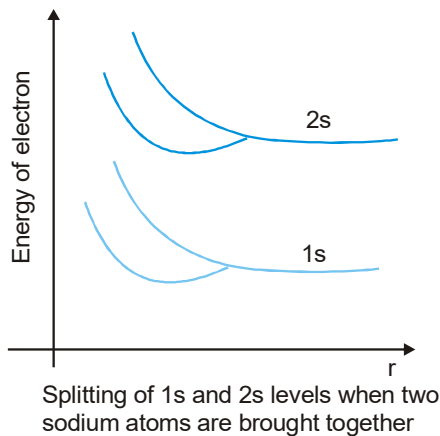


In crystals the number of atoms,  $N$  is very large of the order of  $10^{22}$  to  $10^{23}$  per cubic centimetre, so each energy band contains as many levels as the number of atoms. The spacing between various levels within a band is therefore very small. If for Ex. we assume the total width of a band of energies as 1 eV and  $10^{22}$  levels are to be accommodated within this band, then the average spacing between the adjacent levels is about  $10^{-22}$  eV. For all practical purposes, therefore, energy within a band can be assumed to vary continuously. The formation of bands in a solid is shown schematically in figure(c).

### Energy Bands :

This theory is based on the Pauli exclusion principle.

In isolated atom the valence electrons can exist only in one of the allowed orbitals each of a sharply defined energy called energy levels. But when two atoms are brought nearer to each other, there are alterations in energy levels and they spread in the form of bands.



Energy bands are of following types : –

- (1) **Valence band :** The energy band formed by a series of energy levels containing valence electrons is known as valence band. At 0 K, the electrons fill the energy levels in valence band starting from the lowest one.
  - (i) This band is always filled with electrons.
  - (ii) This is the band of maximum energy.
  - (iii) Electrons are not capable of gaining energy from an external electric field.
  - (iv) No flow of current due to electrons present in this band.
  - (v) The highest energy level which can be occupied by an electron in valence band at 0 K is called the Fermi level.
- (2) **Conduction band :** The higher energy level band is called the conduction band.
  - (i) It is also called an empty band of minimum energy.
  - (ii) This band is partially filled by the electrons.
  - (iii) In this band, the electron can gain energy from an external electric field.
  - (iv) The electrons in the conduction band are called free electrons. They are able to move anywhere within the volume of the solid.
  - (v) Current flows due to such electrons.

- (3) **Forbidden energy gap ( $\Delta E_g$ ) :** Energy gap between conduction band and valence band

$$\Delta E_g = (C.B.)_{\min} - (V.B.)_{\max}$$

## 2. Energy Band Description of Conductor, Insulator and Semiconductor :

The electrical conductivity of materials is a physical quantity which varies over a large span. On one hand we know about metals having very large values of electrical conductivity and on the other hand we have insulators like quartz and mica having negligible conductivity. Beside these there are materials having conductivity (at room temperature) much smaller, than that of metals but much larger than that of insulators these materials are called semiconductors e.g. Silicon and Germanium. Not only that the conductivity of a semiconductor is intermediate, to that of metals and insulators the conductivity of semiconductor varies substantially with temperature. For very low temperature (around 0K) semiconductor behaves like insulator, however, its conductivity increases with increase in temperature.

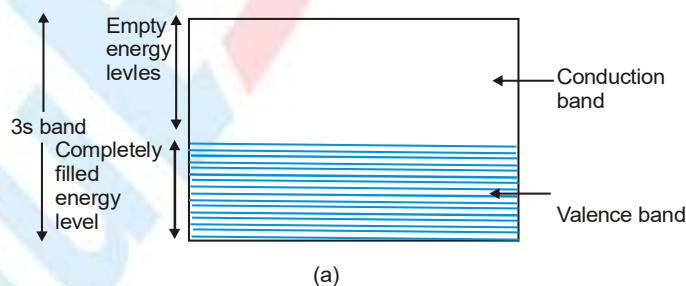
### (a) Conductors :

These are solids in which either the energy band containing valence band is partially filled or the energy band containing valence electrons overlaps with next higher band to give a new band which is partially filled too. For both these situations there are enough free levels available for electrons to which they can be excited by receiving energy from an applied electric field.

### Conduction band and valence band in monovalent metal

Let us consider an Ex. of sodium which is a monovalent metal. Its band structure is such that 1s, 2s and 2p bands are filled with electrons to their capacity however, the 3s band is only half filled. The reason for such a band structure is that for

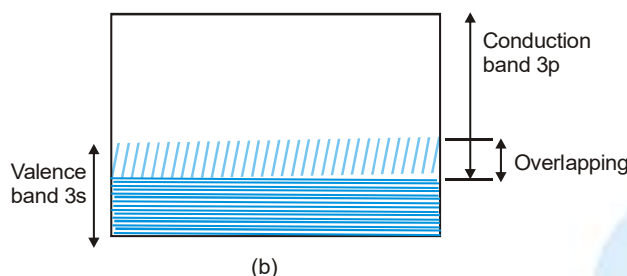
an isolated sodium atom in its electronic structure  $1s^2, 2s^2, 2p^6, 3s^1$  the energy levels 1s, 2s and 2p are filled while 3s contains only one electron against its capacity of accommodating two electrons. The completely filled 1s, 2s and 2p bands do not contribute to electrical conduction because an applied electric field cannot bring about intra band transitions in them. Electrons can also not make band to band transitions from 1s to 2s or from 2s to 2p band as for both these situations unfilled energy levels are not available. However, electrons belonging to 3s band can take part in intra band transitions as half of the energy levels present in this band are available. An applied electric field can impart them an amount of energy sufficient for the transition to free energy levels, and take part in the process of conduction. Thus the conduction properties of sodium are due to this partially filled band which is shown in figure(a). The lower half portion of this band is called valence band and upper half portion is called conduction band as it is in this part when electron reach after receiving energy from electric field the process of conduction starts. All monovalent metals have a half filled conduction band like sodium.



### Conduction band and valence band in bivalent metal

The bivalent elements belonging to the second group of the periodic table e.g magnesium, zinc etc are also metallic. In the solid state of these materials there is an overlapping between the highest filled band and next higher unfilled band. For Ex. magnesium atom (atomic number = 12) has electronic structure -  $1s^2 2s^2 2p^6 3s^2$  and in atomic state there is some energy gap between completely filled 3s level and next higher but unfilled 3p level. However, during the process of crystal formation, the splitting of energy levels take place in such a manner that the 3p band overlaps with 3s band. In the 'hybrid' band so formed now electrons have sufficient number of unfilled levels for transition.

In such situation if 3s band is called valence band then 3p band is conduction band and the two bands overlap as shown in figure (B).



We can conclude that for both the above metals there is no energy gap between maximum energy of valence band and the minimum energy of the conduction band.

The energy that an electron gains from an ordinary current source usually is  $10^{-4}$  to  $10^{-8}$  eV which is sufficient to cause transition between levels inside a partially filled band. As the difference between the adjacent levels is infinitesimal, for such bands the electron can absorb infinitesimal energy in a manner like free electron. Such electrons when reach unfilled higher levels contribute to the process of electric conduction. In metals both the number of free electrons and the vacant energy levels for transitions are very large that is why metals are good conductors of electricity and heat. For metals at ordinary temperature the electrical conductivities are in range  $10^6$  mho/metre to  $10^8$  mho/metre indicating this fact.

## (b) Insulator :

It is a solid in which the energy band formation takes place in such a manner, that the valence band is completely filled while the conduction band is completely empty. In addition to this, these two bands are separated by a large energy gap called forbidden energy gap or band gap. If  $E_c$  and  $E_v$  respectively denotes the minimum energy in conduction band and the maximum energy in valence band then band gap  $E_g$  is defined as

$$E_g = E_c - E_v$$

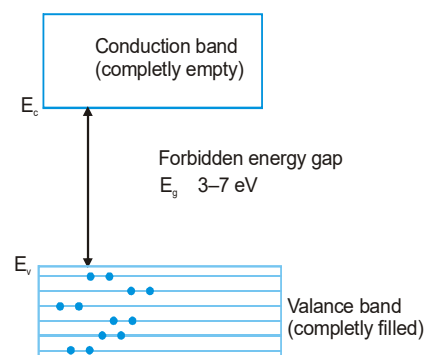
For insulators  $E_g \sim 3$  to 7 eV. As in an empty band no electron is there to

take part in the process of electric conduction, such a band does not contribute in conduction. In a completely filled band very large number of electrons are present but no vacant levels to which these electrons make transition are available and hence again there will not be any conduction. As explained earlier ordinary current sources provide only a very small energy to an electron in a solid and so electrons cannot be excited from valence band to conduction band. Also not only at ordinary temperatures but at elevated temperatures too, the thermal energy is much smaller than the band gap energy  $E_g$  so electrons cannot be excited from valence band to conduction band by thermal means. Consequently solids with such large band gaps

are insulators.

For diamond,  $E_g \approx 6$  eV hence it is insulator.

In general electrical conductivities of insulators are in the range  $10^{-12}$  mho/metre to  $10^{-18}$  mho/metre (resistivity in the range  $10^{12}$  ohm-metre to  $10^{18}$  ohm metre.)





### (c) Semiconductors :

In case of semiconductors, the band structure is essentially of the same type as that for insulators with the only difference that of a relatively smaller forbidden gap. In case of a semiconductor this is typically of the order of 1 eV. At absolute zero temperature, the valence band is completely filled and the conduction band is completely empty and consequently no electrical conduction can result. This is the same behaviour as observed in insulators. i.e at absolute zero a semiconductor behaves like an insulator.

At finite temperatures (room temperature and above) some of the electrons from near the top of valence band acquire enough thermal energy to move into the otherwise empty conduction band. These electrons contribute to the conduction of electricity in a semiconductor.

Also the above said transitions create some unfilled levels in the valence band and the electrons of this band can move into these levels again resulting in conduction. Thus the electrical conductivity of a semiconductor is larger than that of an insulator at room temperature. However since the number of electrons made available to conduction band via this process of thermal excitation is very small as compared to what available for conduction in metals, the conductivity of semiconductors is much smaller than that of metals at a given temperature. Thus the conductivity of semiconductor lies between that of metals and insulators, that is why these are named so. The conductivity of semiconductor increases with temperature.

#### Note : Free electron and Hole in semiconductors.

- (1) When an electron is removed from a covalent bond, it leaves a vacancy behind. An electron from a neighbouring atom can move into this vacancy, leaving the neighbour with a vacancy. In this way the vacancy formed is called hole (or cotter), and can travel through the material and serve as an additional current carriers.
- (2) A hole is considered as a seat of positive charge, having magnitude of charge equal to that of an electron.
- (3) Holes acts as virtual charge, although there is no physical charge on it.
- (4) Effective mass of hole is more than electron.
- (5) Mobility of hole is less than electron.
- (6) Free electron move in CB, while hole in VB in opposite direction.
- (7) Immobile ions are at rest.

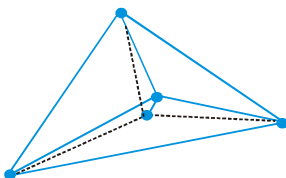
### 3. Intrinsic Semiconductors :

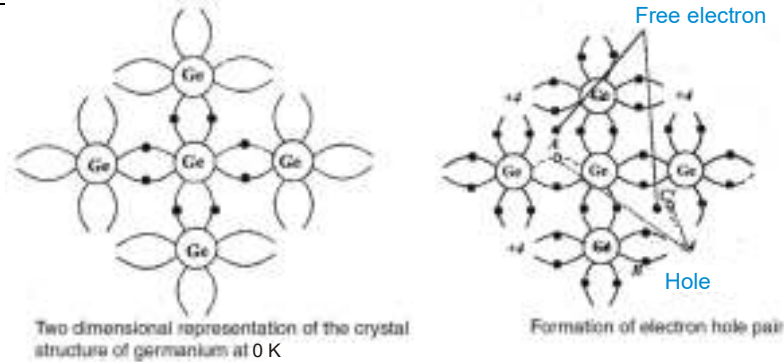
A semiconductor free from impurities is called an intrinsic semiconductor. Ideally an intrinsic semiconductor crystal should contain atoms of this semiconductor only but it is not possible in practice to obtain crystals with such purities. However if the impurity is less than 1 in  $10^8$  part of semiconductor it can be treated as intrinsic. For describing the properties of intrinsic semiconductor we are taking Ex.s of silicon and germanium. Both silicon and germanium are members of the group IV of periodic table of elements and are tetravalent. Their electronic configuration is as follows:

$$\text{Si}(14)=1s^2 2s^2 2p^6 3s^2 3p^2$$

$$\text{Ge}(32)=1s^2 2s^2 2p^6 3s^2 3p^6 3d^{10} 4s^2 4p^2$$

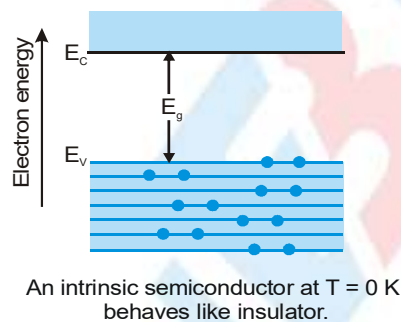
Both elements crystallize in such a way that each atom in the crystal is inside a tetrahedron formed by the four atoms which are closest to it. Figure shows one of these tetrahedral units. Each atom shares its four valence electrons with its immediate neighbours on a one to one basis, so that each atom is involved in four covalent bonds. For convenience, a two dimensional representation of the crystal structure for germanium is shown in figure, which can also be used for silicon (as only covalent bands are being shown).



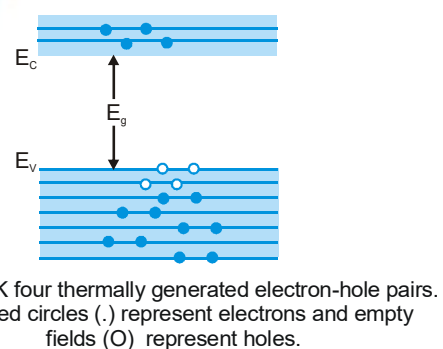


At 0K, all the valence electrons are involved in the bonding and so the crystal is a perfect insulator as there are no free electrons available for conduction. At higher temperatures, however, some of the valence electrons have sufficient energy to break away from the bond and move in the crystal in random manner. Under an applied electric field these electrons drift and conduct electricity.

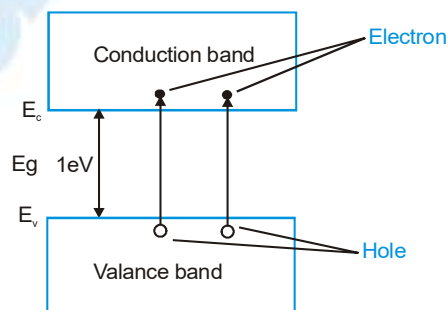
When an electron escapes from a band it leaves behind a vacancy in the lattice. This vacancy is termed as a "hole". The absence of electron amounts to the presence of a positive charge of same magnitude. As explained later, holes also take part in conduction in semiconductors. When a covalent bond is broken, an electron-hole pair is contributed. At room temperature (300K) many electron - hole pairs are present in the crystal. The process of electron - hole generation is explained in figure. Let due to thermal energy an electron is set free from the covalent bond at site A whereby a hole is created at this site. An electron from the covalent bond of a neighbouring atom site B may jump to vacant site A then bond is completed at A but a hole is created at B. In this process a very small energy is involved compared to what is required for an electron - hole pair generation. It is because the electron is jumping from one bond to the other and all electrons in bonding are on an average of same energy. As shown in the figure when an electron jumps from C to B a hole is created at C and so on. In effect then such a vacancy or hole can be considered as mobile. Thus in a semiconductor both electrons and holes act as charge carriers and contribute in electric conduction.



(a)



(b)



The number of electrons and holes generated by thermal means is equal for an intrinsic semiconductor. If  $n_e$  and  $n_h$  represents the electron and hole concentrations respectively then

$$n_i = n_e = n_h$$

$$n_e n_h = n_i^2$$

Here  $n_i$  is intrinsic charge carriers concentration.

- Note :**
- (1) A pure semiconductor is called intrinsic semiconductor. It has thermally generated current carriers.
  - (2) They have four electrons in the outermost orbit of atom and atoms are held together by covalent bond.
  - (3) Free electrons and holes both are charge carriers and  $n_e$  (in C.B.) =  $n_h$  (in V.B.)
  - (4) The drift velocity of electrons ( $v_e$ ) is greater than that of holes ( $v_h$ ).
  - (5) For them fermi energy level lies at the centre of the C.B. and V.B.
  - (6) In pure semiconductor, impurity must be less than 1 in  $10^8$  parts of semiconductor.
  - (7) In intrinsic semiconductor  
 $n_e^{(0)} = n_h^{(0)} = n_i$  ; where  $n_e^{(0)}$  = Electron density in conduction band,  $n_h^{(0)}$  = Hole density in V.B.,  $n_i$  = Density of intrinsic carriers.
  - (8) The fraction of electron of valance band present in conduction band is given by  $f \propto e^{-E_g/kT}$  ; where  $E_g$  = Fermi energy,  $k$  = Boltzmann's constant and  $T$  = Absolute temperature.
  - (9) Because of less number of charge carriers at room temperatre, intrinsic semiconductors have low conductivity so they have no partial use.
  - (10) Number of electrons reaching from valance band to conduction band  $n = AT^{3/2}e^{-E_g/2kT}$  where  $A$  is positive constant.
  - (11) Net charge of a pure semiconductor is zero.

## (a) Electrical conductivity of intrinsic semiconductor:

A semiconductor, at room temperature, contains electrons in the conduction band and holes in the valence band. When an external electric field is applied, the electrons move opposite to the field and the holes move in the direction of the field, thus constituting current in the same direction. The total current is the sum of the electron and hole currents.

$$i = i_e + i_h$$

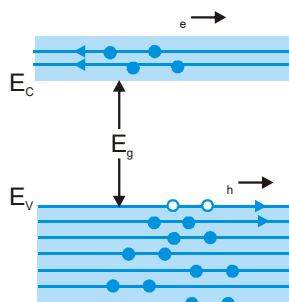
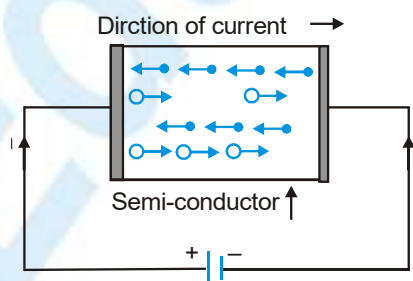
### Hole Current ( $i_h$ ):

Electrons of covalent bond site jump from one position to another position in valency band so hole moves opposite to the jumping of electrons in valence band. Electron originally set free is not involved in the proces of hole motion. Motion of hole is only a convenient way of describing the actual motion of bonded electrons of valence band.

### Electron Current ( $i_e$ )

Free electron moves completely independently as conduction electron and give electron current under the effect of electric field.

It may be noted that apart from the *process of generation* of conduction electrons and holes, a simultaneous *process of recombination* occurs in which the electrons *recombine* with the holes. At equilibrium, the rate of generation is equal to the rate of recombination of charge carriers. The recombination occurs due to an electron colliding with a hole.





Let us consider a semiconductor block of length  $\lambda$ , area of cross-section  $A$  and having electron concentration  $n_e$  and hole concentration is  $n_h$ , across the ends of the semiconductor creates an electric field  $E$  given by

$$E = \frac{V}{\lambda} \quad \text{.....(i)}$$

Under the field  $E$ , the electrons and the holes both drift in opposite directions and constitute currents  $i_e$  and  $i_h$  respectively in the direction of the field. The total current flowing through the semiconductor is

$$i = i_e + i_h$$

If  $v_e$  be the drift velocity of the electrons in the conduction band and  $v_h$  the drift velocity of the holes in the valence band, then we have

$$i_e = n_e e A v_e \quad \text{and} \quad i_h = n_h e A v_h$$

Where  $e$  is the magnitude of electron charge

$$\therefore i = i_e + i_h = eA (n_e v_e + n_h v_h) \quad \text{or} \quad \frac{i}{A} = e (n_e v_e + n_h v_h). \quad \text{.....(ii)}$$

Let  $R$  be the resistance of the semiconductor block and  $\rho$  the resistivity of the block material. Then

$$\rho = RA / \lambda. \quad \text{.....(iii)}$$

Dividing equation (i) by equation (iii), we have

$$\frac{E}{\rho} = \frac{V}{RA} = \frac{i}{A}, \quad (\text{Since } V = iR \text{ by Ohm's law}).$$

Substituting in it the value of  $i/A$  from equation (ii), we get

$$\frac{E}{\rho} = e (n_e v_e + n_h v_h) \quad \text{or} \quad \frac{1}{\rho} = e \left( n_e \frac{v_e}{E} + n_h \frac{v_h}{E} \right). \quad \text{.....(iv)}$$

Let us now introduce a quantity  $\mu$ , called mobility which is defined as the drift velocity per unit field and is expressed in meter<sup>2</sup>/(volt-second). Thus, the mobilities of electron and hole are given by

$$\mu_e = \frac{v_e}{E} \quad \text{and} \quad \mu_h = \frac{v_h}{E}$$

Introducing  $\mu_e$  and  $\mu_h$  in equation (iv), we get

$$\frac{1}{\rho} = e (n_e \mu_e + n_h \mu_h)$$

The electrical conductivity  $\sigma$  is the reciprocal of the resistivity  $\rho$ . Thus, the electrical conductivity of the semiconductor is given by

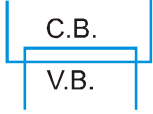
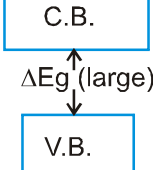
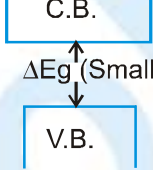
$$\sigma = e (n_e \mu_e + n_h \mu_h) \quad \text{.....(v)}$$

So

$$\sigma = en_i (\mu_e + \mu_h) \quad \rightarrow n_e = n_h = n_i$$

This is the required expression. It shows that the electrical conductivity of a semiconductor depends upon the electron and hole concentrations (number densities) and their mobilities. The electron mobility is higher than the hole mobility.

As temperature rises, both the concentrations  $n_e$  and  $n_h$  increase due to breakage of more covalent bonds. The mobilities  $\mu_e$  and  $\mu_h$ , however, slightly decrease with rise in temperature but this decrease is offset by the much greater increase in  $n_e$  and  $n_h$ . Hence, the conductivity of a semiconductor increases (or the resistivity decreases) with rise in temperature.

Properties	Conductors	Insulators	Semiconductors
Electrical conductivity	$10^2$ to $10^8 \text{ } \Omega^{-1} \text{ m}^{-1}$	$10^{-8} \text{ } \Omega^{-1} \text{ m}^{-1}$	$10^{-5}$ to $10^0 \text{ } \Omega^{-1} \text{ m}^{-1}$
Resistivity	$10^{-2}$ to $10^{-8} \text{ } \Omega \cdot \text{m}$ (negligible)	$10^8 \text{ } \Omega \cdot \text{m}$	$10^5$ to $10^0 \text{ } \Omega \cdot \text{m}$
Band Structure			
Energy gap ( $E_g$ )	Zero or very small	Very large : for diamond it is 6 eV	Ge $\rightarrow$ 0.7 eV Si $\rightarrow$ 1.1 eV GaAs $\rightarrow$ 1.3 eV GaF <sub>2</sub> $\rightarrow$ 2.8 eV
Current carriers	Free electrons	—	Free electrons and holes
Condition of V.B. and C.B. at ordinary temperature	V.B. and C.B. are completely filled or C.B. is some what empty	V.B. – Completely filled C.B. – Completely unfilled	V.B. – some what empty C.B. – some what filled
Temperature co-efficient of resistance	Positive	Zero	Negative
Effect of temperature on conductivity	Decreases	—	Increases
Effect of temperature on resistance	Increases	—	Decreases
Examples	Cu, Ag, Au, Na, Pt, Hg etc.	Wood, plastic, mica, diamond, glass etc.	Ge, Si, GaAs etc.
Electron density	$10^{29} / \text{m}^3$	—	Ge $\sim 10^{19} / \text{m}^3$ Si $\sim 10^{16} / \text{m}^3$

#### 4. Extrinsic semiconductors :

The conductivity of an intrinsic semiconductor depends on its temperature, but at room temperature its conductivity is very low. As such, no important electronic devices can be developed using these semiconductors. Hence there is a necessity of improving their conductivity. This can be done by making use of impurities. When a small amount, say, a few parts per million (ppm), of a suitable impurity is added to the pure semiconductor, the conductivity of the semiconductor is increased manifold. Such materials are known as *extrinsic semiconductors* or *impurity semiconductors*. The deliberate addition of a desirable impurity is called *doping* and the impurity atoms are called *dopants*. Such a material is also called a *doped semiconductor*. The dopant has to be such that it does not distort the original pure semiconductor lattice. It occupies only a very few of the original semiconductor atom sites in the crystal. A necessary condition to attain this is that the sizes of the dopant and the semiconductor atoms should be nearly the same. There are two types of dopants used in doping the tetravalent Si or Ge:

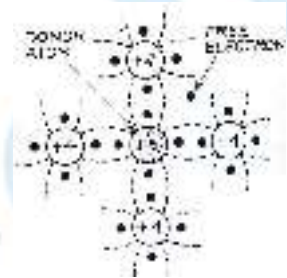
- (i) Pentavalent (valency 5); like Arsenic (As), Antimony (Sb), Phosphorous (P), etc.

- (ii) Trivalent (valency 3); like Indium (In), Boron (B), Aluminium (Al), etc.

We shall now discuss how the doping changes the number of charge carriers (and hence the conductivity) of semiconductors. Si or Ge belongs to the fourth group in the Periodic table and, therefore, we choose the dopant element from nearby fifth or third group, expecting and taking care that the size of the dopant atom is nearly the same as that of Si or Ge. Interestingly, the pentavalent and trivalent dopants in Si or Ge give two entirely different types of semiconductors as discussed below.

**Extrinsic semiconductor are of two types : n-type and p-type.**

- (a) **n-type semiconductor** : When a pentavalent impurity atom (antimony, phosphorus or arsenic) is added to a Ge(or Si) crystal, it replaces a Ge (or Si) atom in the crystal lattice. Four of the five valence electrons of the impurity atom form covalent bonds with one with each valence electron of four Ge (or Si) atoms surrounding.



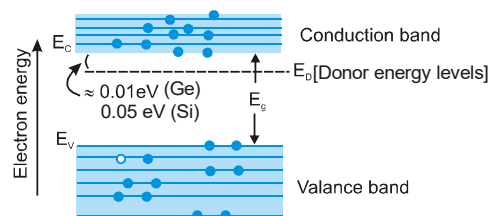
Thus, by adding pentavalent impurity to pure Ge(or Si), the number of free electrons increases, that is, the conductivity of the crystal increases. The impure Ge (or Si) crystal is called an 'n-type' semiconductor because it has an excess of 'negative' charge-carrier (electrons). The impurity atoms are called 'donor' atoms because they donate the conducting electrons to the crystal.

The fifth valence electrons of the impurity atoms occupy some discrete energy levels just below the conduction band. These are called 'donor levels' and are only 0.01 eV below the conduction band in case of Ge, and 0.05 eV below in case of Si. Therefore, at room temperature, the "fifth" electrons of almost all the donor atoms are thermally excited from the donor levels into the conduction band where they move as charge-carriers when an external electric field is applied.

At ordinary temperature, almost all the electrons in the conduction band come from the donor levels, only a few come from the valence band. Therefore, the main charge-carriers responsible for conduction are the electrons contributed by the donors. Since the excitation from the valence band is small, there are very few holes in this band. The current contribution of the holes is therefore small. Thus, in an n-type semiconductor the electrons are the 'majority carriers' and the holes are the 'minority carriers.'

The adjacent figure shows n-type semiconductor at  $T > 0$  K.

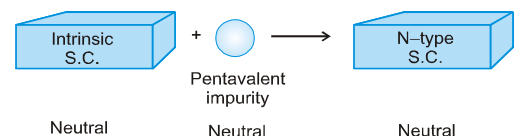
Its conduction band has One thermally generated electron-hole pair and 9 conduction electrons from donor atoms.



## Some important facts about N-Type Semiconductor

These are obtained by adding a small amount of pentavalent impurity to a pure sample of semiconductor (Ge).

- (1) Majority charge carriers – electrons  
Minority charge carriers – hole
- (2)  $n_e \gg n_h ; i_e \gg i_h$
- (3) Conductivity  $\sigma = n_e \mu_e e$
- (4) Donor energy level lies just below the conduction band.



- (5) **Electrons and hole concentration :** In a doped semiconductor, the electron concentration  $n_e$  and the hole concentration  $n_h$  are not equal (as they are in an intrinsic semiconductor). It can be shown that
- $$n_e n_h = n_i^2$$

where  $n_i$  is the intrinsic concentration.

In an n-type semiconductor, the concentration of electrons in conduction band is nearly equal to the concentration of donor atoms ( $N_d$ ) and very large compared to the concentration of holes in valence band. That is :  $n_e \approx N_d \gg n_h$

- (6) Impurity atom called donar atom which is elements of V group of periodic table.  
 (7) Net charge on N type crystal is zero.  
 (8) Immobile charge is positive charge

- (b) **P-Type Semiconductor :** When a trivalent impurity atom (boron, aluminium, gallium or indium) is added to a Ge (or Si) crystal, it also replaces one of the Ge (or Si) atoms in the crystal lattice. Its three valence electrons form covalent bonds with one each valence electron of these Ge (or Si) atoms surrounding it. Thus, there remains an empty space, called a 'p-type' semiconductor because it has an excess of positive 'acceptor' atoms because they create holes which accept electrons.

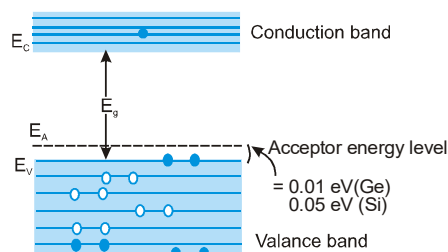


The impurity atoms occupy vacant discrete energy levels just above the top of the valence band. These are called 'acceptor levels'. At room temperature, electrons are easily excited from the valence band into the acceptor levels. The corresponding holes created in the valence band are the main charge-carriers in the crystal when an electric field is applied.

Thus, in a p-type semiconductor the holes are the 'majority carriers' and the few electrons, thermally excited from the valence band into the conduction band, are 'minority carriers'.

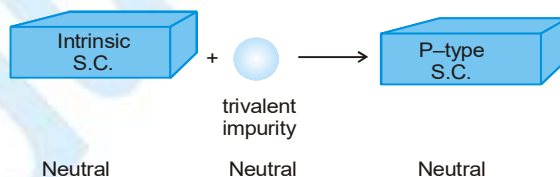
The adjacent figure shows Electron and hole concentration for p-type semiconductor at  $T > 0$  K.

One thermally generated electron-hole pair + seven holes due to acceptor atoms.



### Some Important Facts about P-Type Semiconductor

There are obtained by adding a small amount of trivalent impurity to a pure sample of semiconductor (Ge).



- (1) Majority charge carries – holes  
 Minority charge carries – electrons  
 (2)  $n_h \gg n_e$ ;  $i_h \gg i_e$



- (3) Conductivity  $\sigma \propto n_i \mu_e$
- (4) P-type semiconductor is also electrically neutral (not positively charged)
- (5) Impurity is called Acceptor impurity which is element of III group of the periodic table.
- (6) Acceptor energy level lies just above the valence band.
- (7) **Electron and hole concentration** : In a p-type semiconductor, the concentration of holes in valence band is nearly equal to the concentration of acceptor atoms ( $N_A$ ) and very large compared to the concentration of electron in conduction band. That is  

$$n_2 = N_A \gg n_1$$
- (8) Net charge on p-type crystal is zero.
- (9) Immobile charge is negative charge.

#### Distinction between intrinsic and extrinsic semiconductors :

	Intrinsic Semiconductor		Extrinsic Semiconductor
1	It is a pure nature semiconductor such as pure Ge and pure Si.	1	It is prepared by adding a small quantity of impurity to a pure semiconductor, such as n and p-type semiconductors.
2	In it the concentration of electrons and holes are equal.	2	In it the two concentrations are unequal. There is an excess of electrons in n-type semiconductors and an excess of holes in p-type semiconductors.
3	Its electrical conductivity is very low.	3	Its electrical conductivity is significantly high.
4	Its conductivity cannot be controlled.	4	Its conductivity can be controlled by adjusting the quantity of the impurity added.
5	Its conductivity increases exponentially with temperature.	5	Its conductivity also increases with temperature, but not exponentially.

#### Distinction between n-type and p-type semiconductor :

	N-Type Semiconductor		P-Type Semiconductor
1	It is an extrinsic semiconductor obtained by adding a pentavalent impurity to a pure intrinsic semiconductor.	1	It is an extrinsic semiconductor obtained by adding a trivalent impurity to a pure intrinsic semiconductor.
2	The impurity atoms added provides extra free electrons to the crystal lattice and are called donor atoms.	2	The impurity atoms added create holes in the crystal lattice and are called acceptor atoms because the created holes accept electrons.
3	The electrons are majority carriers and the holes are minority carriers.	3	The holes are majority carriers and the electrons are minority carriers.
4	The electrons concentration is much more than the hole concentration ( $n_e \gg n_h$ ).	4	The hole concentration is much more than the electron concentration ( $n_h \gg n_e$ ).

**Ex.** A silicon specimen is made into a p-type semiconductor by doping on an average one indium atom per  $5 \times 10^7$  silicon atoms. If the number density of atoms in the silicon specimen is  $5 \times 10^{23}$  atoms/m<sup>3</sup>, find the number of acceptor atoms in silicon per cubic centimeter.

**Sol.** The doping of one indium atoms in silicon semiconductor will produce one acceptor atom in p-type semiconductor. Since one indium atom has been doped per  $5 \times 10^7$  silicon atoms, so number density of acceptor atoms

$$\text{in silicon} = \frac{5 \times 10^{23}}{5 \times 10^7} = 10^{16} \text{ atoms/m}^3 = 10^{15} \text{ atoms/cm}^3.$$



## PHYSICS FOR JEE MAIN & ADVANCED

**Ex.** Pure Si at 300K has equal electron ( $n_e$ ) and hole ( $n_h$ ) concentrations of  $1.5 \times 10^{16} \text{ m}^{-3}$ . Doping by indium increases  $n_h$  to  $3 \times 10^{22} \text{ m}^{-3}$ . Calculate  $n_e$  in the doped Si.

**Sol.** For a doped semi-conductor in thermal equilibrium  $n_e n_h = n_i^2$  (Law of mass action)

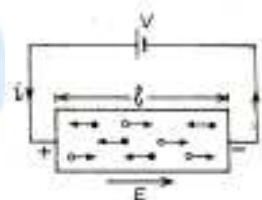
$$n_e = \frac{n_i^2}{n_h} = \frac{(1.5 \times 10^{16})^2}{3 \times 10^{22}} = 7.5 \times 10^9 \text{ m}^{-3}$$

**Ex.** Pure Si at 300 K has equal electron ( $n_e$ ) and hole ( $n_h$ ) concentrations of  $1.5 \times 10^{16} \text{ m}^{-3}$ . Doping by indium increases  $n_h$  to  $4.5 \times 10^{22} \text{ m}^{-3}$ . Calculate  $n_e$  in the doped Si-

**Sol.**  $n_e n_h = n_i^2$   
 $n_h = 4.5 \times 10^{22} \text{ m}^{-3}$   
 So,  $n_e = 5.0 \times 10^9 \text{ m}^{-3}$

### (c) Electrical Conductivity of Extrinsic Semiconductors :

A semiconductor, at room temperature, contains electrons in the conduction band and holes in the valence band. When an external electric field is applied, the electrons move opposite to the field and the hole move in the direction of the field, thus constituting current in the same direction. The total current is the sum of the electron and hole currents.



Let us consider semiconductor block of length  $l$ , area of cross-section  $A$  and having electrons concentration  $n_e$  and hole concentration  $n_h$ . A potential difference  $V$  applied across the ends of the semiconductor creates an electric field  $E$  given by :

$$E = \frac{V}{l} \quad \text{.....(i)}$$

Under the field  $E$ , the electrons and the holes both drift in opposite directions and constitute currents  $i_e$  and  $i_h$  respectively in the direction of the field. The total current flowing through the semiconductor is,

$$i = i_e + i_h$$

If  $v_e$ , be the drift velocity of the electrons in the conduction band and  $v_h$  the drift velocity of the holes in the valence band, then we have

$$i_e = n_e e A v_e \quad \text{and} \quad i_h = n_h e A v_h$$

where  $e$  is the magnitude of electron charge

$$\therefore i = i_e + i_h = eA(n_e v_e + n_h v_h)$$

$$\text{or} \quad \frac{i}{A} = e(n_e v_e + n_h v_h) \quad \text{.....(ii)}$$

Let  $R$  be the resistance of the semiconductor block and  $\rho$  the resistivity of the block material. Then

$$\rho = R A / l \quad \text{.....(iii)}$$

Dividing eq.(i) by eq.(ii) we have

$$\frac{E}{\rho} = \frac{V}{RA} = \frac{i}{A}$$

Because,  $V = iR$  (Ohm's law). Substituting in it the value of  $i/A$  from eq.(ii), we get

$$\frac{E}{\rho} = e(n_e v_e + n_h v_h)$$

$$\text{or} \quad \frac{1}{\rho} = e \left( n_e \frac{v_e}{E} + n_h \frac{v_h}{E} \right) \quad \text{.....(iv)}$$

Let us introduce a quantity  $\mu$ , called mobility which is defined as the drift velocity per unit field and is expressed in  $\text{metre}^2 / (\text{volt}/\text{second})$ . Thus, the mobilities of electrons and hole are given by :

$$\mu_e = \frac{v_e}{E} \quad \text{and} \quad \mu_h = \frac{v_h}{E}$$

Introducing  $\mu_e$  and  $\mu_h$  in eq. (iv), we get

$$\frac{1}{\rho} = e(n_e \mu_e + n_h \mu_h)$$

The electrical conductivity  $\sigma$  is the reciprocal of the resistivity  $\rho$ . Thus, the electrical conductivity of the semiconductor is given by

$$\sigma = e(n_e \mu_e + n_h \mu_h)$$

This is the required expression. It shows that the electrical conductivity of a semiconductor depends upon the electron and hole concentrations (number densities) and their mobilities. The electrons is higher than the hole mobility.

As temperature rises, both the concentration  $n_e$  and  $n_h$  increases due to breakage of more covalent bonds. The mobilities  $\mu_e$  and  $\mu_h$ , however, slightly decrease with rise in temperature but this decrease is offset by the much greater increase in  $n_e$  and  $n_h$ . Hence, the conductivity of a semiconductor increases (or the resistivity decreases) with rise in temperature.

**Ex.** The majority charge carriers in P-type semiconductor are

- (1) Electrons                      (2) Protons                      (3) Holes                      (4) Neutrons

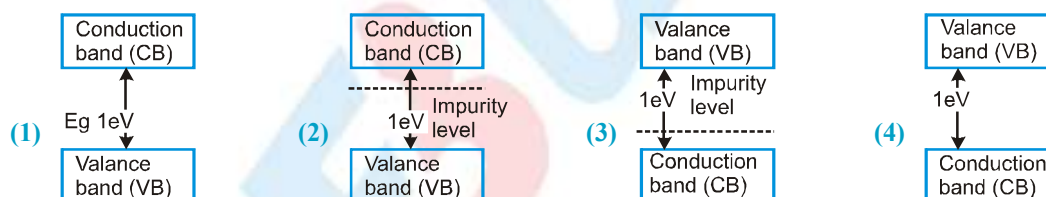
**Sol.** (3) In P-type semiconductors, holes are the majority charge carriers

**Ex.** When a semiconductor is heated, its resistance

- (1) Decreases                      (2) Increases                      (3) Remains unchanged                      (4) Nothing is definite

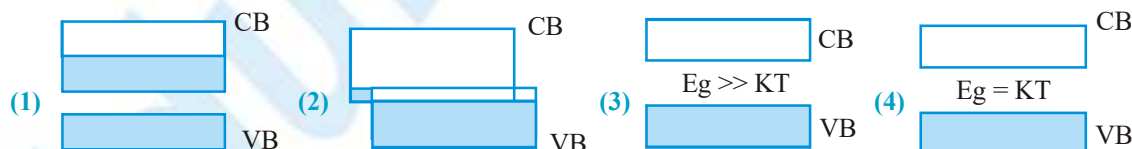
**Ans.** (1)

**Ex.** Which of the following energy band diagram shows the N-type semiconductor



**Sol.** (2) In N-type semiconductor impurity energy level lies just below the conduction band.

**Ex.** Which of the energy band diagram shown in the figure corresponds to that of a semiconductor



**Sol.** (4) In semiconductors, the forbidden energy gap between the valence band and conduction band is very small, almost equal to  $kT$ . Moreover, valence band is completely filled where as conduction band is empty.

**Ex.** The P-N junction is-

- (1) an ohmic resistance                      (2) non ohmic resistance                      (3) a positive resistance                      (4) a negative resistance

**Ans.** (2)

**Ex.** The mean free path of conduction electrons in copper is about  $4 \times 10^{-8}$  m. For a copper block, find the electric field which can give, on an average, 1 eV energy to a conduction electron.

**Sol.** Let the electric field be  $E$ . The force on an electron is  $eE$ . As the electron moves through a distance  $d$ , the work done on it is  $eEd$ . This is equal to the energy transferred to the electron. As the electron travels an average distance of  $4 \times 10^{-8}$  m before a collision, the energy transferred is  $eE(4 \times 10^{-8} \text{ m})$ . To get 1 eV energy from the electric field,

$$eE(4 \times 10^{-8} \text{ m}) = 1 \text{ eV}$$

or,  $E = 2.5 \times 10^7 \text{ V/m.}$

**Ex.** The band gap in germanium is  $\Delta E = 0.68 \text{ eV}$ . Assuming that the number of hole–electron pairs is proportional to  $e^{-\Delta E/2kT}$ , find the percentage increase in the number of charge carries in pure germanium as the temperature is increased from 300 K to 320 K.

**Sol.** The number of charge carries in an intrinsic semiconductor is double the number of hole–electron pairs. If  $N_1$  be the number of charge carries at temperature  $T_1$  and  $N_2$  at  $T_2$ , we have

$$N_1 = N_0 e^{-\Delta E/2kT_1}$$

and  $N_2 = N_0 e^{-\Delta E/2kT_2}$

The percentage increase as the temperature is raised from  $T_1$  to  $T_2$  is

$$f = \frac{N_2 - N_1}{N_1} \times 100 = \left( \frac{N_2}{N_1} - 1 \right) \times 100 = 100 \left[ e^{\frac{\Delta E}{2k} \left( \frac{1}{T_1} - \frac{1}{T_2} \right)} - 1 \right]$$

Now  $\frac{\Delta E}{2k} \left( \frac{1}{T_1} - \frac{1}{T_2} \right) = \frac{0.68 \text{ eV}}{2 \times 8.62 \times 10^{-5} \text{ eV/K}} \left( \frac{1}{300 \text{ K}} - \frac{1}{320 \text{ K}} \right) = 0.82$

Thus  $f = 100 \times [e^{0.82} - 1] \approx 127$ .

Thus, the number of charge carries increase by about 127%.

**Ex.** The energy of a photon of sodium light ( $\lambda = 589 \text{ nm}$ ) equals the band gap of a semiconducting material. (a) Find the minimum energy  $E$  required to create a hole–electron pair. (b) Find the value of  $E/kT$  at a temperature of 300 K.

**Sol.** (a) The energy of the photon is  $E = \frac{hc}{\lambda}$

$$= \frac{1242 \text{ eV} \cdot \text{nm}}{589 \text{ nm}} = 2.1 \text{ eV.}$$

Thus the band gap is 2.1 eV. This is also the minimum energy  $E$  required to push an electron from the valence band into the conduction band. Hence, the minimum energy required to create a hole–electron pair is 2.1 eV. So it is difficult for the thermal energy to create the hole–electron pair but a photon of light can do it easily.

(b) At  $T = 300 \text{ K}$ ,

$$kT = (8.62 \times 10^{-5} \text{ eV/K})(300 \text{ K})$$

$$= 25.86 \times 10^{-3} \text{ eV.}$$

Thus,  $\frac{E}{kT} = \frac{2.1 \text{ eV}}{25.86 \times 10^{-3} \text{ eV}} = 81$ .



## 5. Junction Diode

A junction diode is a basic semiconductor device. It is a semiconductor crystal having acceptor impurities in one region (P – type crystal) and donor impurities in the other region (n-type crystal). The boundary between the two regions is called ‘p–n junction’.

### Circuit Symbol for a p-n Junction Diode

In electronic circuits, the semiconductor devices are represented by their symbols. The symbol for the basic device, the p-n junction diode, is shown in Fig. The arrow-head represents the p -region and the bar represents the n -region of the diode. The direction of the arrow is from p to n and indicates the direction of conventional current flow under forward bias. The p -side is called 'anode' and the n -side is called 'cathode'.

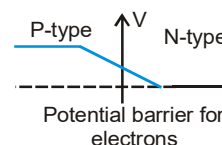
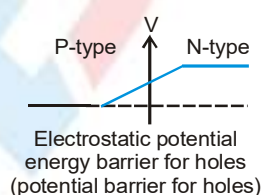
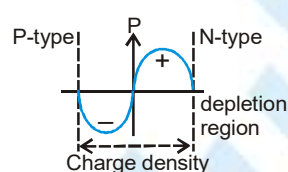
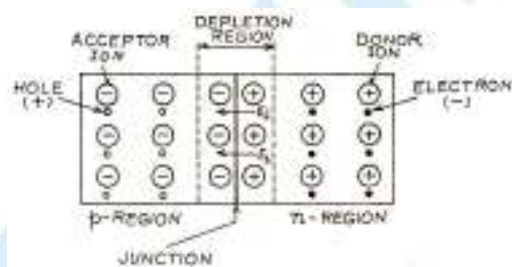


#### (a) Formation of p-n Junction :

A p-n junction is not the interface between p -type and n – type semiconductor crystals pressed together. It is a single piece of semiconductor crystal having an excess of acceptor impurities into one side and of donor impurities into the other.

#### (b) Potential Barrier at the Junction: Formation of Depletion Region:

A p-n junction is shown in Fig. The p -type region has (positive) holes as majority charge-carriers, and an equal number of fixed negatively-charged acceptor ions. (The material as a whole is thus neutral). Similarly, the n -type region has (negative) electrons as majority charge-carriers, and an equal number of fixed positively-charged donor ions.



The region on either of the junction which becomes depleted (free) of the mobile charge-carriers is called the 'depletion region'. The width of the depletion region is of the order of  $10^{-6}$  m. The potential difference developed across the depletion region is called the 'potential barrier'. It is about 0.3 volt for Ge, p-n junction and about 0.7 volt for silicon p-n junction. It, however, depends upon the dopant concentration in the semiconductor.

The magnitude of the barrier electric field for a silicon junction is

$$E_i \approx \frac{V}{d} \approx \frac{0.7}{10^{-6}} = 7 \times 10^5 \text{ Vm}^{-1}$$

## Diffusion & Drift Current

Due to concentration difference hole try to diffuse from p side to n side but due to depletion layer only those hole are able to diffuse from p to n side which have high kinetic energy. Similarly electron of high kinetic energy also diffuse from n to p so diffusion current flow from p to n side.

Due to thermal collision or increase in temperature some valence electron comes in conduction band. If this occurs in depletion region then hole move towards p side & electron move towards n side so a current produce from n to p side. It is called drift current, in steady state both diffusion & drift current are equal & opposite.

**Ex.** In a p-n junction with open ends,

- (1) there is no systematic motion of charge carriers
- (2) holes and conductor electrons systematically go from the p-side and from the n-side to the p-side respectively
- (3) there is no net charge transfer between the two sides
- (4) there is a constant electric field near the junction

**Ans.** (3,4)

**Ex.** A potential barrier of 0.50 V exists across a P-N junction. If the depletion region is  $5.0 \times 10^{-7}$  m wide, the intensity of the electric field in this region is

- (1)  $1.0 \times 10^6$  V/m
- (2)  $1.0 \times 10^5$  V/m
- (3)  $2.0 \times 10^5$  V/m
- (4)  $2.0 \times 10^6$  V/m

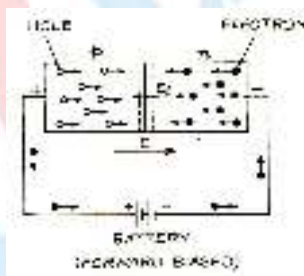
**Sol.** (1)  $E = \frac{V}{d} = \frac{0.5}{5 \times 10^{-7}} = 10^6$  V/m

### (c) Forward and Reverse Biasing of Junction Diode

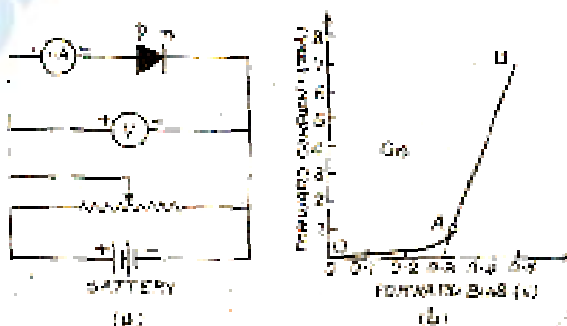
The junction diode can be connected to an external battery in two ways, called 'forward biasing' and 'reverse biasing' of the diode. It means the way of connecting emf source to P-N junction diode. It is of following two types:

#### (i) Forward Biasing :

A junction diode is said to be forward-biased when the positive terminal of the external battery is connected to the p -region and the negative terminal to the n -region of the diode.



**Forward-Biased Characteristics :** The circuit connections are shown in Fig. . The positive terminal of the battery is connected to the p -region and the negative terminal to the n -region of the



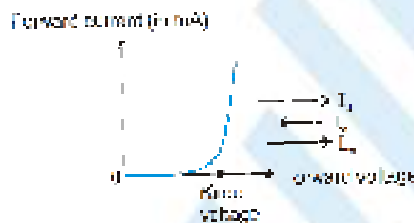


junction diode through a potential-divider arrangement which enables to change the applied voltage. The voltage is read by a voltmeter  $V$  and the current by a milliammeter  $mA$ . Starting with a low value, the forward bias voltage is increased step by step and the corresponding forward current is noted. A graph is then plotted between voltage and current. The resulting curve  $OAB$  (Fig. b) is the forward characteristic of the diode.

In the beginning, when the applied voltage is low, the current through the junction diode is almost zero. It is because of the potential barrier (about  $0.3\text{ V}$  for Ge p-n junction and about  $0.7\text{ V}$  for Si junction) which opposes the applied voltage. With increase in applied voltage, the current increases very slowly and non-linearly until the applied voltage exceeds the potential barrier. This is represented by the portion  $OA$  of the characteristic curve. With further increase in applied voltage, the current increases very rapidly and almost linearly. Now the diode behaves as an ordinary conductor. This is represented by the straight-line part  $AB$  of the characteristic. If this straight line is projected back, it intersects the voltage-axis at the barrier potential voltage.

**Note :**

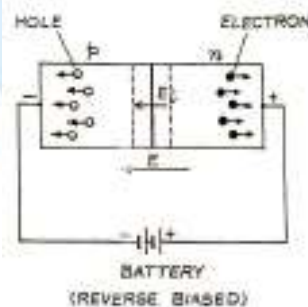
- (i) In forward biasing width of depletion layer decreases
- (ii) In forward biasing resistance offered  $R_{\text{Forward}} \approx 10\Omega - 25\Omega$
- (iii) Forward bias opposes the potential barrier and for  $V > V_B$  a forward current is set up across the junction.
- (iv) Cut-in (Knee) voltage : The voltage at which the current starts to increase rapidly. For Ge it is  $0.3\text{ V}$  and for Si it is  $0.7\text{ V}$ .



- (v)  $I_{df}$  –diffusion current,  $I_{dr}$  –drift current

**(ii) Reverse Biasing:**

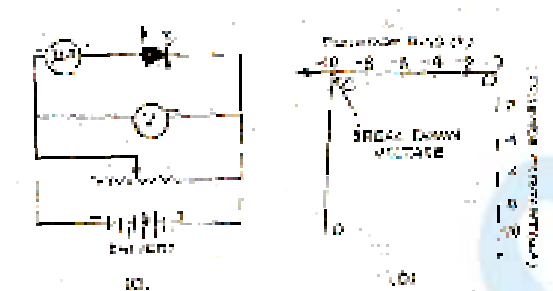
A junction diode is said to be reverse-biased when the positive terminal of the external battery is connected to the  $n$ -region and the negative terminal to the  $p$ -region of the diode (Fig.)



In this condition, the external field  $E$  is directed from  $n$  toward  $p$  and thus aids the internal barrier field  $E_i$ . Hence holes in the  $p$ -region and electrons in the  $n$ -region are both pushed away from the junction, that is, they cannot combine at the junction. Thus, there is almost no current due to flow of majority carriers.

## Reverse-Biased Characteristic:

The circuit connections are shown in Fig. (a) in which the positive terminal of the battery is connected to the n - region and the negative terminal to the p -region of the junction diode. In reverse-biased diode, a very small current (of the order of  $\mu\text{A}$ ) flows across the junction due to the motion of the few thermally-generated minority-carriers (electrons in p -region and holes in n -region) whose motion is aided by the applied voltage. The small reverse current remains almost constant over a sufficiently long range of reverse bias (applied voltage), increasing very little with increasing bias. This is represented by the part OC of the reverse characteristic curve (Fig b).



### Note:

- (i) In reverse biasing width of depletion layer increases
- (ii) In reverse biasing resistance offered  $R_{\text{Reverse}} \approx 10^5 \Omega$
- (iii) Reverse bias supports the potential barrier and no current flows across the junction due to the diffusion of the majority carriers.
- (A very small reverse currents may exist in the circuit due to the drifting of minority carriers across the junction)
- (iv) Break down voltage : Reverse voltage at which break down of semiconductor occurs. For Ge it is 25V and for Si it is 35 V.

## (d) Avalanche Breakdown :

The avalanche breakdown occurs in lightly doped junction. If the reverse bias is made very high, the minority-carriers acquire kinetic energy enough to break the covalent bonds near the junction, thus liberating electron-hole pairs. These charge-carriers are accelerated and produce, in the same way, other electron-hole pairs. The process is cumulative and an avalanche of electron-hole pairs is produced. The reverse current then increases abruptly to a relatively large value (part CD of the characteristic). This is known as 'avalanche breakdown' and may damage the junction by the excessive heat generated. The reverse bias voltage at which the reverse current increase abruptly is called the 'breakdown voltage'.

## Zener Breakdown

Zener breakdown occurs in heavily doped junctions. Under a high reverse - bias voltage, the p-n junction's depletion region expands, leading to a high strength electric field across the junction. A sufficiently strong electric field manages to break the covalent bonds of the semiconductor atoms, which liberates a large number of free minority carries. The sudden generation of carries rapidly increases the reverse current and gives rise to high slope resistance of Zener diode.

The reverse bias voltage at which the reverse current increase abruptly is called the 'ZENER breakdown voltage' or 'Zener voltage'. The numerical value of the breakdown voltage varies from tens of volts to several hundred volts depending on the number density of the impurity atoms doped into the diode.

(e) **Dynamic Resistance of a Junction Diode**

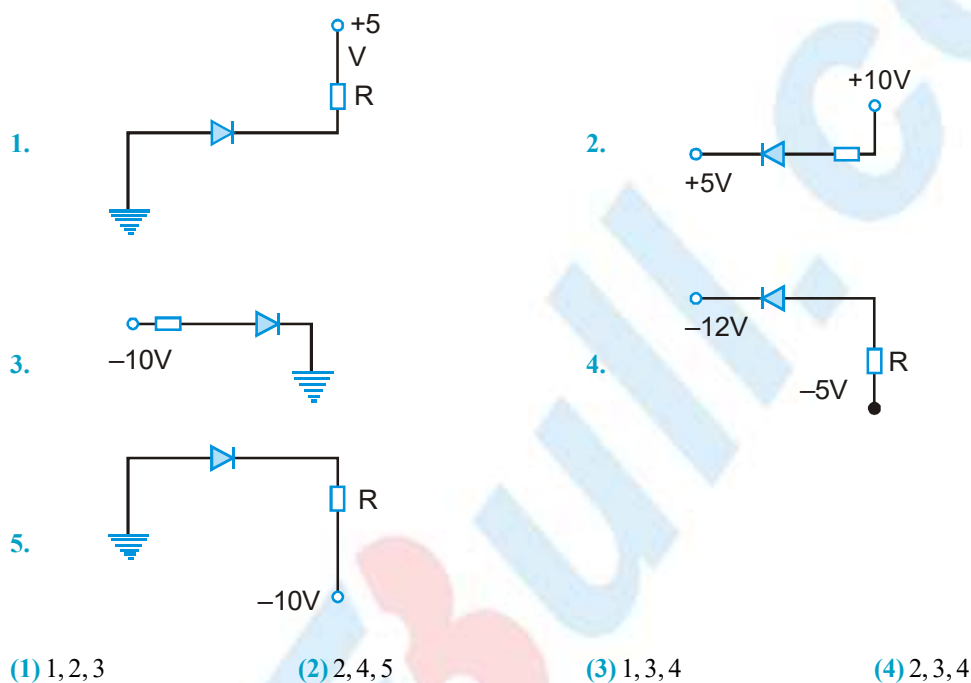
The current-voltage curve of junction diode shows that the current does not vary linearly with the voltage, that is, Ohm's law is not obeyed. In such situation, a quantity known as 'dynamic resistance' (or a.c. resistance) is defined.

The dynamic resistance of a junction diode is defined as the ratio of a small change in applied voltage ( $\Delta V$ ) to the corresponding small change in current ( $\Delta i$ ), that is

$$R_d = \frac{\Delta V}{\Delta i}$$

In the forward characteristic of p-n junction diode, beyond the turning point (knee), however, the current varies almost linearly with voltage. In this region,  $R_d$  is almost independent of  $V$  and Ohm's law is obeyed.

**Ex.** In the given figure, which of the diodes are forward biased ?



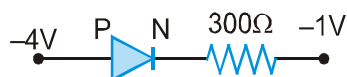
**Sol.** (2) In figure 2, 4 and 5. P-crystals are more positive as compared to N-crystals.

**Ex.** Two identical capacitors A and B are charged to the same potential  $V$  and are connected in two circuits at  $t = 0$  as shown in fig. The charges on the capacitor at a time  $t = CR$  are, respectively,



**Ans.** 2

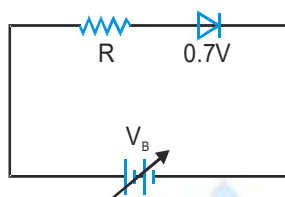
**Ex.** What is the current in the circuit shown below



- (1) 0 amp      (2)  $10^{-2}$  amp      (3) 1 amp      (4) 0.10 amp

**Sol.** (1) The potential of P-side is more negative than that of N-side, hence diode is in reverse biasing. In reverse biasing it acts as an open circuit, hence no current flows.

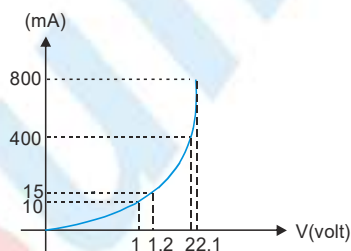
**Ex.** Assume that the junction diode in the following circuit requires a minimum current of 1 mA to be above the knee point (0.7V) of its I-V characteristic curve. Also assume that the voltage across the diode is independent of current above the knee point. If  $V_B = 5V$ , what should be the maximum value of R so that the voltage is above the knee point-



- (1) 4.3 kΩ      (2) 860 kΩ      (3) 4.3 Ω      (4) 860 Ω

**Ans.** (1)

**Ex.** The i-V characteristic of a p-n junction diode is shown in figure. Find the approximate dynamic resistance of the p-n junction when (a) a forward bias of 1 volt is applied, (b) a forward bias of 2 volt is applied



(a) The current at 1 volt is 10 mA and at 1.2 volt it is 15 mA. The dynamic resistance in this region is

$$R = \frac{\Delta V}{\Delta i} = \frac{0.2 \text{ volt}}{5 \text{ mA}} = 40 \Omega$$

(b) The current at 2 volt is 400 mA and at 2.1 volt it is 800 mA. The dynamic resistance in the region is

$$R = \frac{\Delta V}{\Delta i} = \frac{0.1 \text{ volt}}{400 \text{ mA}} = 0.25 \Omega.$$

## 6. p-n Junction Diode as a Rectifier

An electronic device which converts alternating current / voltage into direct current / voltage is called 'rectifier'.

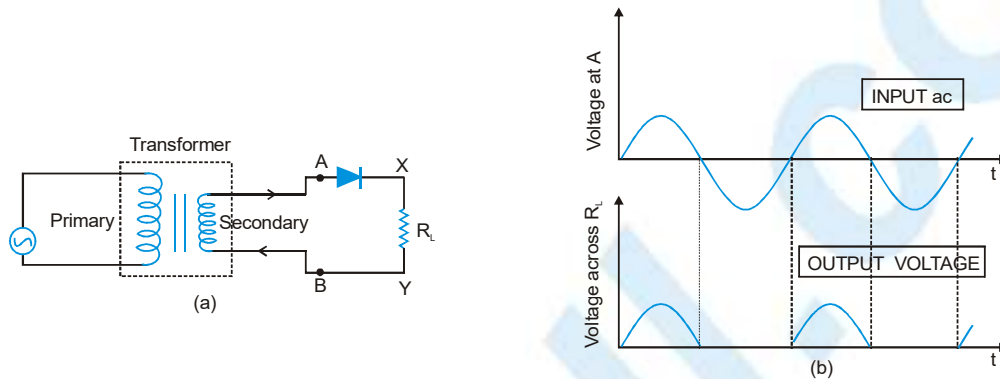
A p-n junction diode offers a low resistance for the current to flow, when forward-biased, but a very high resistance, when reverse-biased. It thus passes current only in one direction and acts as a rectifier.

The junction diode can be used either as a half-wave rectifier, when it allows current only during the positive half-cycles of the input a.c. supply; or as a full-wave rectifier when it allows current in the same direction for both half-cycles of the input alternating current.

(a) **p-n Junction Diode as Half-wave Rectifier**

The half-wave rectifier circuit is shown in Fig. (a) and the input and output wave forms in Fig. (b). The alternating input voltage is applied across the primary  $P_1P_2$  of a transformer.  $S_1S_2$  is the secondary coil of the same transformer.  $S_1$  is connected to the p-type crystal of the junction diode and  $S_2$  is connected to the n-type crystal through a load resistance  $R_L$ .

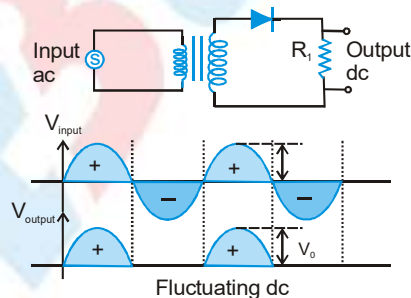
During the first half-cycle of the a.c. input, when the terminal  $S_1$  of the secondary is positive and  $S_2$  is negative, the junction diode is forward-biased. Hence it conducts and current flows through the load  $R_L$  in the direction shown by arrows. The current produces across the load an output voltage of the same shape as the half-cycle of the input voltage. During the second half-cycle of the a.c. input, the terminal  $S_1$  is negative and  $S_2$  is positive. The diode is now reverse-biased. Hence there is almost zero current and zero output voltage across  $R_L$ . The process is repeated. Thus, the output current is unidirectional, but pulsating, as shown in lower part of Fig. (b).



Since the output- current corresponds to one half of the input voltage wave, the other half being missing, the process is called half-wave rectification.

The purpose of the transformer is to supply the necessary voltage to the rectifier. If direct current at high voltage is to be obtained from the rectifier, as is necessary for power supply, then a step-up transformer is used, as shown in Fig. (a). In many solid-state equipments, however, direct current of low voltage is required. In that case, a step-down transformer is used in the rectifier.

**Note:**



- (i) During positive half cycle  
Diode  $\rightarrow$  forward biased  
Output signal  $\rightarrow$  obtained
- (ii) During negative half cycle  
Diode  $\rightarrow$  reverse biased  
Output signal  $\rightarrow$  not obtained
- (iii) Output voltage is obtained across the load resistance  $R_L$ . It is not constant but pulsating (mixture of ac and dc) in nature.



(iv) Average output in one cycle

$$I_{dc} = \frac{I_0}{\pi} \text{ and } V_{dc} = \frac{V_0}{\pi}; I_0 = \frac{V_0}{r_f + R_L}$$

( $r_f$  = forward biased resistance)

(v) r.m.s. output :  $I_{rms} = \frac{I_0}{2}$ ,  $V_{rms} = \frac{V_0}{2}$

(vi) The ratio of the effective alternating component of the output voltage or current to the dc component is known as ripple factor.

$$r = \frac{I_{ac}}{I_{dc}} = \left[ \left( \frac{I_{rms}}{I_{dc}} \right)^2 - 1 \right]^{1/2} = 1.21$$

(vii) Peak inverse voltage (PIV) : The maximum reverse biased voltage that can be applied before commencement of Zener region is called the PIV. When diode is not conducting PIV across it =  $V_0$

(viii) Efficiency : It is given by  $\% \eta = \frac{P_{out}}{P_{in}} \times 100 = \frac{40.6}{1 + \frac{r_f}{R_L}}$

If  $R_L \gg r_f$  then  $\eta = 40.6 \%$

If  $R_L = r_f$  then  $\eta = 20.3 \%$

(ix) Form factor =  $\frac{I_{rms}}{I_{dc}} = \frac{\pi}{2} = 1.57$

(x) The ripple frequency ( $\omega$ ) for half wave rectifier is same as that of ac.

(b) **p-n Junction Diode as Full-wave Rectifier:** In a full-wave rectifier, a unidirectional, pulsating output current is obtained for both halves of the a.c. input voltage. Essentially, it requires two junction diodes so connected that one diode rectifies one half and the second diode rectifies the second half of the input.

The circuit for a full-wave rectifier is shown in Fig. 8 (a) and the input and output wave forms in Fig. (b). The a.c. input voltage is applied across the primary  $P_1P_2$  of a transformer. The terminals  $S_1$  and  $S_2$  of the secondary are connected to the p-type crystals of the junction diodes  $D_1$  and  $D_2$  whose n-type crystals are connected to each other. A load resistance  $R_L$  is connected across the n-type crystals and the central-tap T of the secondary  $S_1S_2$ .

During the first half-cycle of the a.c. input voltage, the terminal  $S_1$  is suppose positive relative to T and  $S_2$  is negative. In this situation, the junction diode  $D_1$  is forward-biased and  $D_2$  is reverse-biased. Therefore,  $D_1$  conducts while  $D_2$  does not. The conventional current flows through diode  $D_1$ , load  $R_L$  and the upper half of the secondary winding, as shown by solid arrows. During the second half-cycle of the input voltage,  $S_1$  is negative relative to T and  $S_2$  is positive. Now,  $D_1$  is reverse-biased and does not conduct while  $D_2$  is forward-biased and conducts. The current now flows through  $D_2$ , load  $R_L$  and the lower half of the secondary, as shown by dotted arrows. It may be seen that the current in the load  $R_L$  flows in the same direction for both half-cycles of the a.c. input voltage. Thus, the output current is a continuous series of unidirectional pulses. However, it can be made fairly steady by means of smoothing filters.

**Note :**

(i) During positive half cycle

Diode  $\longrightarrow$  forward biased

Output signal  $\longrightarrow$  obtained

(ii) During negative half cycle

Diode  $\longrightarrow$  reverse biased

Output signal  $\longrightarrow$  not obtained

(iii) Output voltage is obtained across the load resistance  $R_L$ . It is not constant but pulsating (mixture of ac and dc) in nature.

(iv) Average output in one cycle

$$I_{dc} = \frac{I_0}{\pi} \text{ and } V_{dc} = \frac{V_0}{\pi}; I_0 = \frac{V_0}{r_f + R_L}$$

( $r_f$  = forward biased resistance)

(v) r.m.s. output :  $I_{rms} = \frac{I_0}{2}$ ,  $V_{rms} = \frac{V_0}{2}$

(vi) The ratio of the effective alternating component of the output voltage or current to the dc component is known as ripple factor.

$$r = \frac{I_{ac}}{I_{dc}} = \left[ \left( \frac{I_{rms}}{I_{dc}} \right)^2 - 1 \right]^{1/2} = 1.21$$

(vii) Peak inverse voltage (PIV) : The maximum reverse biased voltage that can be applied before commencement of Zener region is called the PIV. When diode is not conducting PIV across it =  $V_0$

(viii) Efficiency :  $\eta\% = \frac{P_{out}}{P_{in}} \times 100 = \frac{40.6}{1 + \frac{r_f}{R_L}}$

If  $R_L \gg r_f$  then  $\eta = 40.6\%$

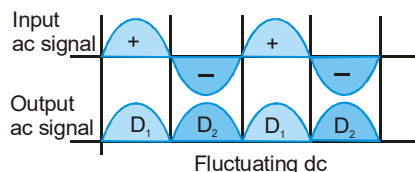
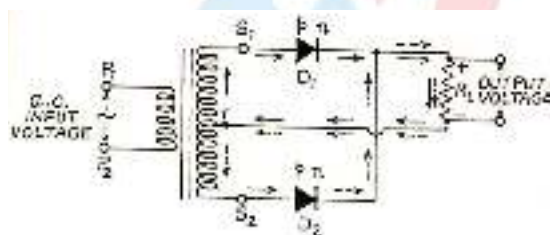
If  $R_L = r_f$  then  $\eta = 20.3\%$

(ix) Form factor =  $\frac{I_{rms}}{I_{dc}} = \frac{\pi}{2} = 1.57$

(x) The ripple frequency ( $\omega$ ) for half wave rectifier is same as that of ac.

(b) **p-n Junction Diode as Full-wave Rectifier:** In a full-wave rectifier, a unidirectional, pulsating output current is obtained for both halves of the alternating input voltage. Essentially, it requires two junction diodes so connected that one diode rectifies one half and the second diode rectifies the second half of the input.

The circuit for a full-wave rectifier is shown in Fig. (a) and the input and output wave forms in Fig. (b). The alternating input voltage is applied across the primary  $P_1P_2$  of a transformer. The terminals  $S_1$  and  $S_2$  of the secondary are connected to the p-type crystals of the junction diodes  $D_1$  and  $D_2$  whose n-type crystals are connected to each other. A load resistance  $R_L$  is connected across the n-type crystals and the central-tap T of the secondary  $S_1S_2$ .

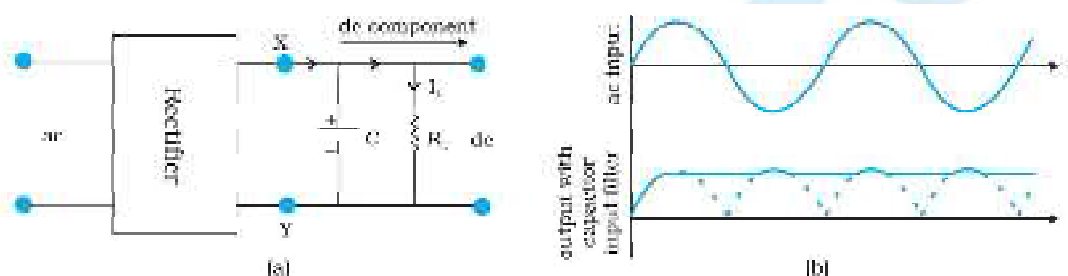


During the first half-cycle of the a.c. input voltage, the terminal  $S_1$  is suppose positive relative to T and  $S_2$  is negative. In this situation, the junction diode  $D_1$  is forward-biased and  $D_2$  is reverse-biased. Therefore,  $D_1$  conducts while  $D_2$  does not. The conventional current flows through diode  $D_1$ , load  $R_L$  and the upper half of the secondary winding, as shown by solid arrows. During the second half-cycle of the input voltage,  $S_1$  is negative relative to T and  $S_2$  is positive. Now,  $D_1$  is reverse-biased and does not conduct while  $D_2$  is forward-biased and conducts. The current now flows through  $D_2$ , load  $R_L$  and the lower half of the secondary, as shown by dotted arrows. It may be seen that the current in the load  $R_L$  flows in the same direction for both half-cycles of the alternating input voltage. Thus, the output current is a continuous series of unidirectional pulses. However, it can be made fairly steady by means of smoothing filters.

## Filter

The rectified voltage is in the form of pulses of the shape of half sinusoids. Though it is unidirectional it does not have a steady value. To get steady dc output from the pulsating voltage normally a capacitor is connected across the output terminals (parallel to the load  $R_L$ ). One can also use an inductor in series with  $R_L$  for the same purpose. Since these additional circuits appear to filter out the ac ripple and give a pure dc voltage, so they are called filters.

Now we shall discuss the role of capacitor in filtering. When the voltage across the capacitor is rising, it gets charged. If there is no external load, it remains charged to the peak voltage of the rectified output. When there is a load, it gets discharged through the load and the voltage across it begins to fall. In the next half-cycle of rectified output it again gets charged to the peak value. The rate of fall of the voltage across the capacitor depends upon the inverse product of capacitor  $C$  and the effective resistance  $R_L$  used in the circuit and is called the time constant. To make the time constant large value of  $C$  should be large. So capacitor input filters use large capacitors. The output voltage obtained by using capacitor input filter is nearer to the peak voltage of the rectified voltage. This type of filter is most widely used in power supplies.

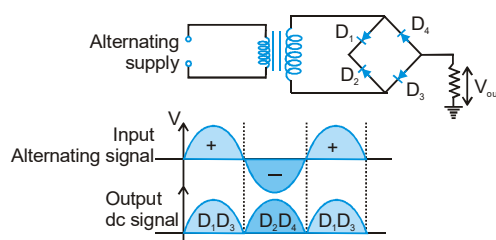


**FIGURE** (a) A full-wave rectifier with capacitor filter, (b) Input and output voltage of rectifier in (a).

- (i) During positive half cycle  
 Diode :  $D_1 \longrightarrow$  forward biased  $D_2 \longrightarrow$  reverse biased  
 Output signal  $\longrightarrow$  obtained due to  $D_1$  only
- (ii) During negative half cycle  
 Diode :  $D_1 \longrightarrow$  reverse biased  $D_2 \longrightarrow$  forward biased  
 Output signal  $\longrightarrow$  obtained due to  $D_2$  only
- (iii) Fluctuating dc  $\longrightarrow$  Filter  $\longrightarrow$  constant dc.
- (iv) Output voltage is obtained across the load resistance  $R_L$ . It is not constant but pulsating in nature.
- (v) Average output :  $V_{av} = \frac{2V_0}{\pi}$ ,  $I_{av} = \frac{2I_0}{\pi}$
- (vi) r.m.s. output  $= V_{rms} = \frac{V_0}{\sqrt{2}}$ ,  $I_{rms} = \frac{I_0}{\sqrt{2}}$
- (vii) Ripple factor :  $r = 0.48 = 48\%$
- (viii) Ripple frequency : The ripple frequency of full wave rectifier  $= 2 \times$  (Frequency of input ac)
- (ix) Peak inverse voltage (PIV) : It's value is  $2V_0$ .
- (x) Efficiency :  $\eta = \frac{81.2}{1 + \frac{r_f}{R_L}}$  for  $r_f \ll R_L$ ,  $\eta = 81.2\%$

(3) **Full wave bridge rectifier :** Four diodes  $D_1$ ,  $D_2$ ,  $D_3$  and  $D_4$  are used in the circuit.

During positive half cycle  $D_1$  and  $D_3$  are forward biased and  $D_2$  and  $D_4$  are reverse biased. During negative half cycle  $D_2$  and  $D_4$  are forward biased and  $D_1$  and  $D_3$  are reverse biased.



(c) **Different Types of Junction Diode**

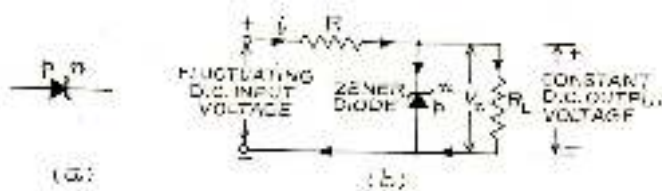
The junction diodes are of many types. The important types are Zener diode, photodiode, light-emitting diode (LED) and solar cell.

(i) **Zener Diode:** It is a voltage-regulating device based upon the phenomenon of avalanche breakdown in a junction diode.

When the reverse-bias applied to a junction diode is increased, there is an abrupt rise in the (reverse) current when the bias reverse reaches a certain value, known as 'breakdown voltage' or 'Zener voltage'.

Thus, in this region of the reverse characteristic curve, the voltage across the diode remains almost constant for a large range of currents. Hence the diode may be used to stabilize voltage at a pre-determined value. It is then known as 'Zener diode'. It can be designed, by properly controlled doping of the diode, to stabilize voltage at any desired value between 4 –100 volt. .

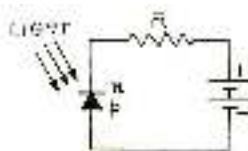
Fig. (a) shows the symbol of a Zener diode and Fig. (b) shows a simple circuit for stabilizing voltage across a load  $R_L$ . The circuit consists of a series voltage-dropping resistance  $R$  and a Zener diode in



parallel with the load  $R_L$ . The Zener diode is selected with a Zener voltage  $V_Z$  equal to the voltage desired across the load. The fluctuating d.c. input voltage may be the d.c. output of a rectifier. Whenever the input voltage increases, the excess voltage is dropped across the resistance  $R$ . This causes an increase in the input current  $i$ . This increase is conducted by the Zener diode, while the current through the load and hence the voltage across it remains constant at  $V_Z$ . Likewise, a decrease in the input voltage causes a decrease in the input current  $i$ . The current through the diode decreases correspondingly, again maintaining the current through the load constant.

Since the resistance  $R$  absorbs the input voltage fluctuations to give a constant output voltage  $V_Z$ , the circuit cannot work if the input voltage falls below  $V_Z$ .

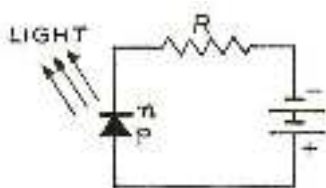
(ii) **Photodiode :** A photodiode is a reverse-biased p-n junction made from a photosensitive semiconductor. The junction is embedded in clear plastic. The upper surface across the junction is open to light, while the remaining sides of the plastic are painted black or enclosed in a metallic case. The entire unit is extremely small, of the order of a 0.1 inch size.



When no light is falling on the junction and the reverse-bias is of the order of a few tenths of a volt, an almost constant small current ( $\approx \mu\text{A}$ ) is obtained. This "dark" current is the reverse saturation current due to the thermally-generated minority-carriers (electrons in p-region and holes in n-region). When light of appropriate frequency is made incident on the junction, additional electron-hole pairs are created near the junction (due to breaking of covalent bonds). These light-generated minority-carriers cross the (reverse-biased) junction and contribute to the (reverse) current due to thermally-generated carriers. Therefore, the current in the circuit increases (a fraction of a mA). This, so-called 'photoconductive' current varies almost linearly with the incident light flux.

The p-n photodiodes can operate at frequencies of the order of 1 MHz. Hence they are used in high-speed reading of computer punched cards, light-detection systems, light-operated switches, electronic counters etc.

**(iii) Light-Emitting Diode (LED) :** When a p-n junction diode is forward-biased, both the electron and the holes move towards the junction. As they cross the junction, the electrons fall into the holes (recombine). Hence, energy is released at the junction (because the electrons fall from a higher to a lower energy level). In case of Ge and Si diodes, the energy released is infra-red radiation. If, however, the diode is made of gallium arsenide or indium phosphide, the energy released is visible light. The diode is then called a 'light-emitting diode' (LED).



LEDs have replaced incandescent lamps in many applications because of their low input power, long life and fast on-off switching.

They are extensively used in fancy electronic devices like calculators, etc.

**Ex.** A zener diode of voltage  $V_Z (= 6 \text{ Volt})$  is used to maintain a constant voltage across a load resistance  $R_L (= 1000\Omega)$  by using a series resistance  $R_S (= 100\Omega)$ . If the e.m.f. or source is  $E (= 9\text{V})$ , calculate the value of current through series resistance, Zener diode and load resistance. What is the power being dissipated in Zener diode.

**Sol.** Here,  $E = 9\text{V}$ ;  $V_Z = 6$ ;  $R_L = 1000\Omega$  and  $R_S = 100\Omega$ ,

Potential drop across series resistor  $V = E - V_Z = 9 - 6 = 3 \text{ V}$

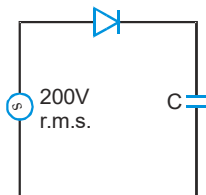
Current through series resistance  $R_S$  is  $I = \frac{V}{R} = \frac{3}{100} = 0.03\text{A}$

Current through load resistance  $R_L$  is  $I_L = \frac{V_Z}{R_L} = \frac{6}{1000} = 0.006\text{A}$

Current through Zener diode is  $I_Z = I - I_L = 0.03 - 0.006 = 0.024 \text{ A}$

Power dissipated in Zener diode is  $P_Z = V_Z I_Z = 6 \times 0.024 = 0.144 \text{ Watt}$

**Ex.** In the figure, an A.C. of rms voltage 200 volt is applied to the circuit containing diode and the capacitor and it is being rectified. The potential across the capacitor C in volt will be-



(1) 500

(2) 200

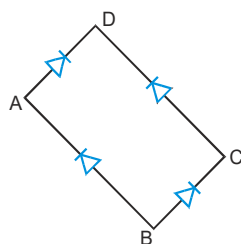
(3) 283

(4) 141

**Ans.** (3)



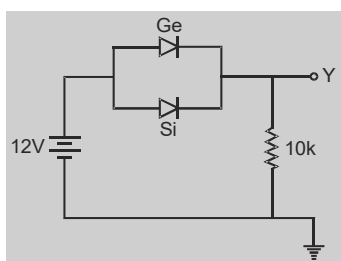
**Ex.** In the figure, input is applied across A and C and output is taken across B and D, then the output is-



- (1) Zero                      (2) Same as input                      (3) Full wave rectified                      (4) Half wave rectified

**Ans.** (3)

**Ex.** Two junction diodes one of germanium (Ge) and other of silicon (Si) are connected as shown in figure to a battery of emf 12 V and a load resistance 10 kΩ. The germanium diode conducts at 0.3 V and silicon diode at 0.7 V. When a current flows in the circuit, the potential of terminal Y will be-



- (1) 12 V                      (2) 11 V                      (3) 11.3 V                      (4) 11.7 V

**Ans.** (4)

**Ex.** Potential barrier developed in a junction diode opposes-

- (1) Minority carriers in both regions only                      (2) Majority carriers  
(3) Electrons in N-region                      (4) Holes in P-region

**Ans.** (2)

**Ex.** Avalanche breakdown in a semiconductor diode occurs when-

- (1) Forward current exceeds a certain value                      (2) Reverse bias exceeds a certain value  
(3) Forward bias exceeds a certain value                      (4) The potential barrier is reduced to zero

**Ans.** (2)

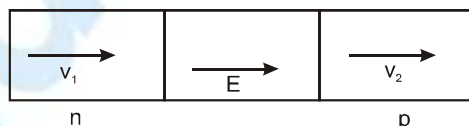
**Ex.** A potential barrier of 0.50 V exists across a p-n junction.

- (a) If the depletion region is  $5.0 \times 10^{-7}$  m wide, what is the intensity of the electric field in this region ?  
(b) An electron with speed  $5.0 \times 10^5$  m/s approaches the p-n junction from the n-side. With what speed will it enter the p-side ?

**Sol. :** (a) The electric field is  $E = V/d$

$$= \frac{0.50 \text{ V}}{5.0 \times 10^{-7} \text{ m}} = 1.0 \times 10^6 \text{ V/m.}$$

(b)



Suppose the electron has a speed  $v_1$  when it enters the depletion layer and  $v_2$  when it comes out of it (figure). As the potential energy increases by  $e \times 0.50$  V, from the principle of conservation of energy,

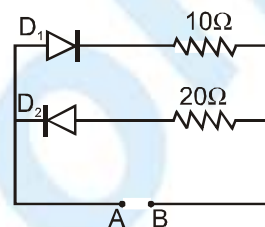
$$\frac{1}{2}mv_1^2 = e \times 0.50 \text{ V} + \frac{1}{2}mv_2^2 \quad \text{or,}$$

$$\frac{1}{2} \times (9.1 \times 10^{-31} \text{ kg}) \times (5.0 \times 10^5 \text{ m/s})^2 = 1.6 \times 10^{-19} \times 0.5 \text{ J} + \frac{1}{2} (9.1 \times 10^{-31} \text{ kg}) v_2^2$$

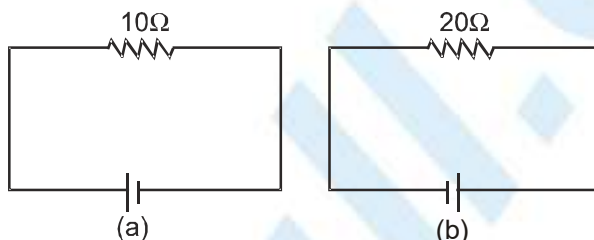
$$\text{or, } 1.13 \times 10^{-19} \text{ J} = 0.8 \times 10^{-19} \text{ J} + (4.55 \times 10^{-31} \text{ kg}) v_2^2.$$

$$\text{Solving this, } v_2 = 2.7 \times 10^5 \text{ m/s}$$

**Ex.** A 2 V battery may be connected across the points A and B as shown in figure. Assume that the resistance of each diode is zero in forward bias and infinity in reverse bias. Find the current supplied by the battery if the positive terminal of the battery is connected to (a) the point A (B) the point B.

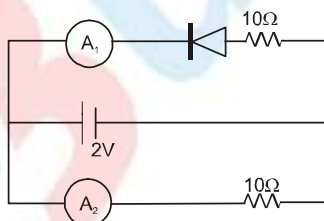


**Sol. :** (a) When the positive terminal of the battery is connected to the point A, the diode  $D_1$  is forward-biased and  $D_2$  is reverse-biased. The resistance of the diode  $D_1$  is zero, and it can be replaced by a resistanceless wire. Similarly, the resistance of the diode  $D_2$  is infinity, and it can be replaced by a broken wire. The equivalent circuit is shown in figure. The current supplied by the battery is  $2 \text{ V}/10 \Omega = 0.2 \text{ A}$ .



(b) When the positive terminal of the battery is connected to the point B, the diode  $D_2$  is forward-biased and  $D_1$  is reverse biased. The equivalent circuit is shown in figure (b). The current through the battery is  $2 \text{ V}/20 \Omega = 0.1 \text{ A}$ .

**Ex.** What are the reading of the ammeters  $A_1$  and  $A_2$  shown in figure. Neglect the resistance of the meters.



**Ans.** Reading of  $A_1$  is zero, Reading of  $A_2$  is 0.2 A

**Ex.** Calculate the value of  $V_0$ , if the Si diode and the Ge diode start conducting at 0.7 V and 0.3 V respectively, in the given circuit. If the Ge diode connection be reversed, what will be the new values of  $V_0$  and  $I$ ?

**Sol.** The effective forward voltage across Ge diode is  $12 \text{ V} - 0.3 = 11.7$ .

This will appear as the output voltage across the load, that is,

$$V_0 = 11.7 \text{ V}$$

The current in the load is



$$i = \frac{V_0}{R_L} = \frac{11.7}{5k\Omega} = 2.34 \text{ mA.}$$

On reversing the connections of Ge diode, it will be reverse-biased and conduct no current. Only Si diode will conduct. The effective forward voltage across Si diode is  $12 \text{ V} - 0.7 \text{ V} = 11.3 \text{ V}$ . This will appear as output, that is

$$V_0 = 11.3 \text{ V}$$

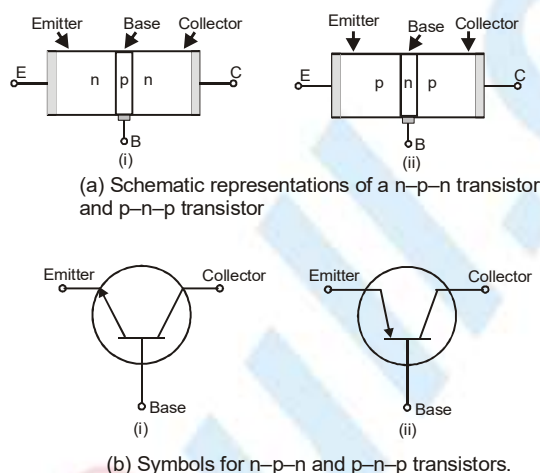
The current in the load is

$$i = \frac{V_0}{R_L} = \frac{11.3}{5k\Omega} = 2.26 \text{ mA.}$$

## 7. Junction Transistor :

### Transistor structure and action :

A transistor has three doped regions forming two p-n junctions between them. There are two types of transistors, as shown in figure.



**(i) n-p-n transistor :** Here two segments of n-type semiconductor (emitter and collector) are separated by a segment of p-type semiconductor (base).

**(ii) p-n-p transistor :** Here two segments of p-type semiconductor (termed as emitter and collector) are separated by a segment of n-type semiconductor (termed as base).

The schematic representations of an n-p-n and a p-n-p configuration are shown in figure. All the three segments of a transistor have different thickness and their doping levels are also different. In the schematic symbols used for representing p-n-p and n-p-n transistors (figure b) the arrowhead shows the direction of conventional current in the transistor. A brief description of the three segments of a transistors is given below:

**Emitter :** This is the segment on one side of the transistor shown in fig.(a). It is of moderate size and heavily doped. It supplies a large number of majority carriers for the current flow through the transistor.

**Base :** This is the central segment. It is very thin and lightly doped.

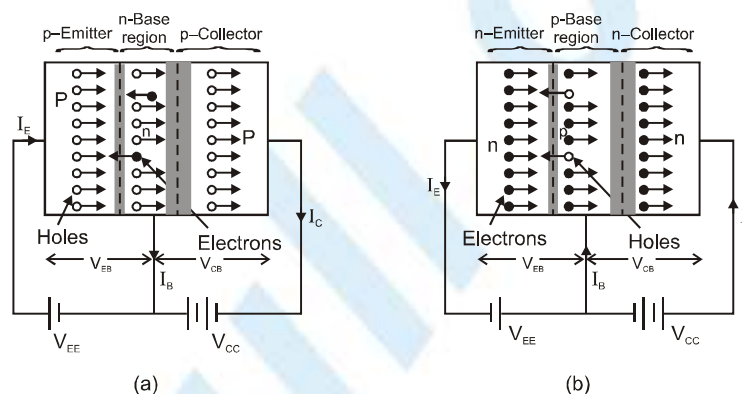
**Collector :** This segment collects a major portion of the majority carriers supplied by the emitter. The collector side is moderately doped and larger in size as compared to the emitter, so that heat generated during the collection of charge carriers may be easily dissipated into atmosphere

In case of a p-n junction, there is a formation of depletion region across the junction. In case of a transistor, there are two depletion regions are formed respectively at the emitter base-junction and the base collector junction.

The transistor works as an amplifier, with its emitter-base junction forward biased and the base-collector junction

reverse biased. This situation is shown in figure, where  $V_{CC}$  and  $V_{EE}$  are used for creating the respective biasing. When the transistor is biased in this way it is said to be in active state. We represent the voltage between emitter and base as  $V_{EB}$  and that between the collector and base as  $V_{CB}$ . In figure, base is a common terminal for the two power supplies whose other terminals are connected to emitter and collector, respectively. So, the two power supplies are represented as  $V_{EE}$  and  $V_{CC}$  respectively. In circuits, where emitter is the common terminal, the power supply between the base and emitter is represented as  $V_{BB}$  and that between collector and emitter as  $V_{CC}$ .

The heavily doped emitter has a high concentration of majority carriers, which will be holes in a p-n-p transistor and electrons in an n-p-n transistor. These majority carriers enter the base region in large numbers. The base is thin and lightly doped. So, the majority carriers there would be few. In a p-n-p transistor the majority carriers in the base are electrons since base is of n-type semiconductor. The large number of holes entering the base from the emitter swamps the small number of electrons there. As the base collector-junction is reverse biased, these holes, which appear as minority carriers at the junction, can easily cross the junction and enter the collector. The holes in the base could move either towards the base terminal to combine with the electrons entering from outside or cross the junction to enter into the collector and reach the collector terminal. The base is made thin so that most of the holes find themselves near the reverse-biased base-collector junction and so cross the junction instead of moving to the base terminal.



Bias Voltage applied on : (a) p-n-p transistor and (b) n-p-n transistor

**Note :** Due to forward bias a large current enters the emitter-base junction, but most of it is diverted to adjacent reverse-biased base-collector junction and the current coming out of the base becomes a very small fraction of the current that entered the junction. If we represent the hole current and the electron current crossing the forward biased junction by the sum  $I_h + I_e$ . We see that the emitter current  $I_E = I_h + I_e$  but the base current  $I_B \ll I_h + I_e$ , because a major part of  $I_E$  goes to collector instead of coming out of the base terminal. The base current is thus a small fraction of the emitter current.

It is obvious from the above description and also from a straight forward application of Kirchoff's law to figure(a) that the emitter current is the sum of collector current and base current :

$$I_E = I_C + I_B$$

We also see that  $I_C \approx I_E$ .

Our description of the direction of motion of the holes is identical with the direction of the conventional current. But the direction of motion of electrons is just opposite to that of the current. Thus in a p-n-p transistor the current enters from emitter into base whereas in a n-p-n transistor it enters from the base into the emitter. The arrowhead in the emitter shows the direction of the conventional current.

We can conclude that in the active state of the transistor the emitter-base junction acts as a low resistance while the base collector acts as a high resistance.

In a transistor, only three terminals are available viz emitter (E), base (B) and collector (C). Therefore in a circuit the input/output connections have to be such that one of these (E,B or C) is common to both the input and the output. Accordingly, the transistor can be connected in either of the following three configurations :

**Common Emitter (CE), Common Base (CB), Common Collector (CC).**

### Working of Transistor

- (1) There are four possible ways of biasing the two P-N junctions (emitter junction and collector junction) of transistor.
- (i) Active mode : Also known as linear mode operation.
- (ii) Saturation mode : Maximum collector current flows and transistor acts as a closed switch from collector to emitter terminals.
- (iii) Cut-off mode : Denotes operation like an open switch where only leakage current flows.
- (iv) Inverse mode : The emitter and collector are inter changed.

### Different modes of operation of a transistor

Operating mode	Emitter base bias	Collector base bias
Active	Forward	Reverse
Saturation	Forward	Forward
Cut off	Reverse	Reverse
Inverse	Reverse	Forward

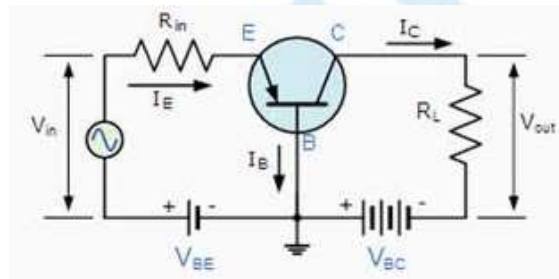
- (2) A transistor is mostly used in the active region of operation i.e., emitter base junction is forward biased and collector base junction is reverse biased.
- (3) From the operation of junction transistor it is found that when the current in emitter circuit changes. There is corresponding change in collector current.
- (4) In each state of the transistor there is an input port and an output port. In general each electrical quantity (V or I) obtained at the output is controlled by the input.

### Transistor Configurations

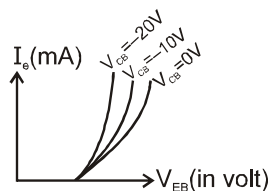
A transistor can be connected in a circuit in the following three different configurations.

Common base (CB), Common emitter (CE) and Common collector (CC) configuration.

- (1) **CB configurations** : Base is common to both emitter and collector.



- (i) Input current =  $I_e$  (ii) Input voltage =  $V_{EB}$  (iii) Output voltage =  $V_{CB}$  (iv) Output current =  $I_c$
- With small increase in emitter-base voltage  $V_{EB}$ , the emitter current  $I_e$  increases rapidly due to small input resistance.
- (v) **Input characteristics** : If  $V_{CB} = \text{constant}$ , curve between  $I_e$  and  $V_{EB}$  is known as input characteristics. It is also known as emitter characteristics :



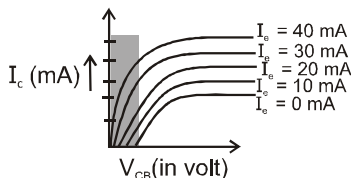


Input characteristics of NPN transistor are also similar to the above figure but  $I_c$  and  $V_{EB}$  both are negative and  $V_{CB}$  is positive. Dynamic input resistance of a transistor is given by

$$R_i = \left( \frac{\Delta V_{EB}}{\Delta I_e} \right)_{V_{CB} = \text{constant}} \quad \{R_i \text{ is of the order of } 100\Omega\}$$

(vi) **Output characteristics :** Taking the emitter current  $i_e$  constant, the curve drawn between  $I_c$  and  $V_{CB}$  are known as output characteristics of CB configuration.

$$\text{Dynamics output resistance } R_o = \left( \frac{\Delta V_{CB}}{\Delta i_c} \right)_{i_e = \text{constant}}$$



**Note: Transistor as CB amplifier**

(i) ac current gain  $\alpha_c = \frac{\text{Small change in collector current } (\Delta i_c)}{\text{Small change in emitter current } (\Delta i_e)}$

(ii) dc current gain  $\alpha_{dc}$  (or  $\alpha$ ) =  $\frac{\text{Collector current } (i_c)}{\text{Emitter current } (i_e)}$

value of  $\alpha_{dc}$  lies between 0.95 to 0.99

(iii) Voltage gain  $A_v = \frac{\text{Change in output voltage } (\Delta V_o)}{\text{Change in input voltage } (\Delta V_i)}$

$\Rightarrow A_v = \alpha_{ac} \times \text{Resistance gain}$

(iv) Power gain =  $\frac{\text{Change in output power } (\Delta P_o)}{\text{Change in input power } (\Delta P_i)}$

$\Rightarrow \text{Power gain} = \alpha_{ac}^2 \times \text{Resistance gain}$

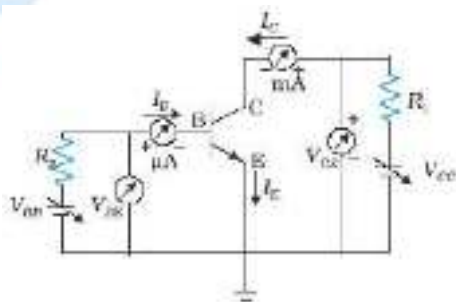
**Common Emitter(CE) :**

The transistor is most widely used in the CE configuration.

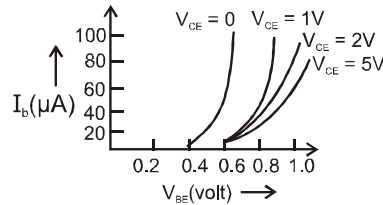
When a transistor is used in CE configuration, the input is between the base and the emitter and the output is between the collector and the emitter. The variation of the base current  $I_B$  with the base-emitter voltage  $V_{BE}$  is called the input characteristic. The output characteristics are controlled by the input characteristics. This implies that the collector current changes with the base current.

**CE configurations :** Emitter is common to both base and collector.

The graphs between voltages and currents when emitter of a transistor is common to input and output circuits are known as CE characteristics of a transistor.

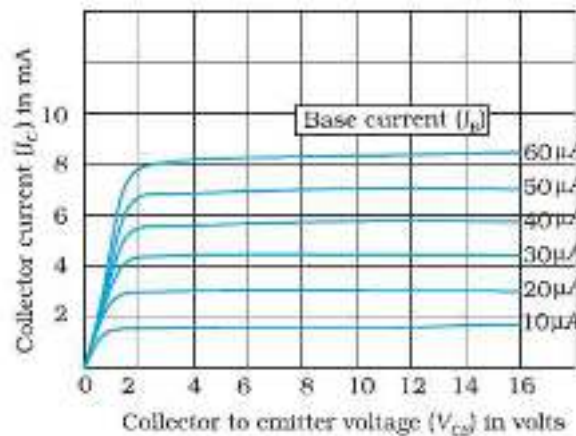


**Input characteristics :** Input characteristics curve is drawn between base current  $I_B$  and emitter base voltage  $V_{EB}$ , at constant collector emitter voltage  $V_{CE}$ .



$$\text{Dynamic input resistance } R_i = \left( \frac{\Delta V_{BE}}{\Delta I_B} \right)_{V_{CE} \rightarrow \text{constant}}$$

**Output characteristics :** Variation of collector current  $I_C$  with  $V_{CE}$  can be noticed for  $V_{CE}$  between 0 to 1 V only. The value of  $V_{CE}$  up to which the  $I_C$  changes with  $V_{CE}$  is called knee voltage. The transistor are operated in the region above knee voltage.



$$\text{Dynamic output resistance } R_o = \left( \frac{\Delta V_{CE}}{\Delta I_C} \right)_{I_B \rightarrow \text{constant}}$$

**(b) Transistor as a device :**

The transistor can be used as a device application depending on the configuration used (namely CB, CC and CE), the biasing of the E-B and B-C junction and the operation region namely cutoff, active region and saturation.

When the transistor is used in the cutoff or saturation state it acts as a switch. On the other hand for using the transistor as an amplifier, it has to operate in the active region.

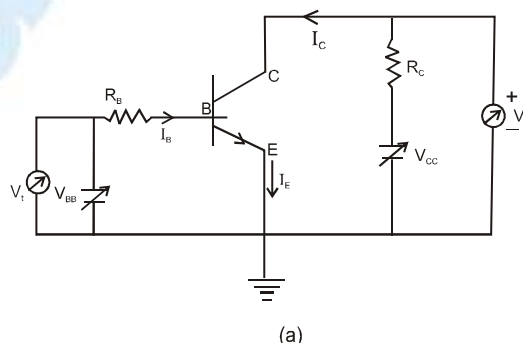
**(i) Transistor as a switch:**

We shall try to understand the operation of the transistor as a switch by analysing the behaviour of the base-biased transistor in CE configuration as shown in fig. (a). Applying Kirchhoff's voltage rule to the input and output sides of this circuit, we get

$$V_{BB} = I_B R_B + V_{BE}$$

and

$$V_{CE} = V_{CC} - I_C R_C$$

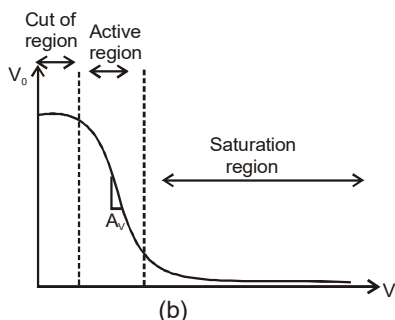


We shall treat  $V_{BB}$  as the dc input voltage  $V_i$  and  $V_{CE}$  as the dc output voltage  $V_o$ . So, we have

$$V_i = I_B R_B + V_{BE}$$

and

$$V_o = V_{CC} - I_C R_C$$



Let us see how  $V_o$  changes as  $V_i$  increases from zero onwards. In the case of Si transistor, as long as input  $V_i$  is less than 0.6 V, the transistor will be in cut off state and current  $I_C$  will be zero. Hence  $V_o = V_{CC}$ .

When  $V_i$  becomes greater than 0.6 V the transistor is in active state with some current  $I_C$  in the output path and the output  $V_o$  decreases as the term  $I_C R_C$  increases. With increase of  $V_i$ ,  $I_C$  increases almost linearly and so  $V_o$  decreases linearly till its value becomes less than about 1.0 V.

Beyond this, the change becomes non linear and transistor goes into saturation state. With further increase in  $V_i$  the output voltage is found to decrease further towards zero though it may never become zero. If we plot the  $V_o$  vs  $V_i$  curve, [also called the transfer characteristics of the base-biased transistor (figure b)], we see that between cut off state and active state and also between active state and saturation state there are regions of non-linearity showing that the transition from cutoff state to active state and from active state to saturation state are not sharply defined.

As long as  $V_i$  is low and unable to forward-bias the transistor,  $V_o$  is high (at  $V_{CC}$ ). If  $V_i$  is high enough to drive the transistor into saturation very near to zero. When the transistor is not conducting it is said to be switched off and when it is driven into saturation it is said to be switched on. This shows that if we define low and high states as below and above certain voltage levels corresponding to cutoff and saturation of the transistor, then we can say that a low input switches the transistor off and a high input switches it on.

**(ii) Transistor as an Amplifier (CE-Configuration) :** To operate the transistor as an amplifier it is necessary to fix its operating point somewhere in the middle of its active region. If we fix the value of  $V_{BB}$  corresponding to a point in the middle of the linear part of the transfer curve then the dc base current  $I_B$  would be constant and corresponding collector current  $I_C$  will be constant. The dc voltage  $V_{CE} = V_{CC} - I_C R_C$  would also remain constant. The operating values of  $V_{CE}$  and  $I_B$  determine the operating point, of the amplifier. If a small sinusoidal voltage with amplitude  $v_s$  is superposed on the dc base bias by connecting the source of that signal in series with the  $V_{BB}$  supply, then the base current will have sinusoidal variations superimposed on the value of  $I_B$ . As a consequence the collector current also will have sinusoidal variations superimposed on the value of  $I_C$  producing in turn corresponding change in the value of  $V_o$ . We can measure the ac variations across the input and output terminals by blocking the dc voltages by larger capacitors.

In the discription of the amplifier given above we have not considered any ac signal. In general, amplifiers are used to amplify alternating signals. Now let us superimpose an ac input signal  $v_i$  (to be amplified) on the bias  $V_{BB}$  (dc) as shown in Figure. The output is taken between the collector and the ground.

The working of an amplifier can be easily understood, if we first assume that  $v_i = 0$ . Then applying Kirchhoff's law to the output loop, we get

$$V_{CC} = V_{CE} + I_C R_L$$

Likewise, the input loop gives

$$V_{BB} = V_{BE} + I_B R_B$$

when  $v_i$  is not zero, we get

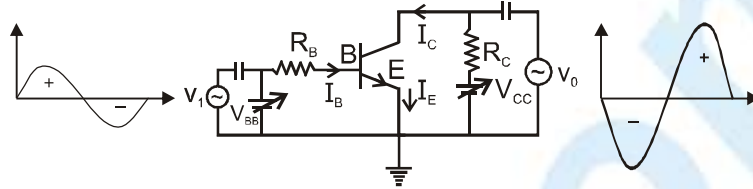
$$V_{BE} + v_i = V_{BE} + I_B R_B + \Delta I_B (R_B + r_i)$$

The change in  $V_{BE}$  can be related to the input resistance  $r_i$  and the change in  $I_B$ .

Hence 
$$v_i = \Delta I_B (R_B + r_i)$$
$$= r_i \Delta I_B$$

The change in  $I_B$  causes a change in  $I_C$ . We define a parameter  $\beta_{ac}$ , which is similar to the  $\beta_{dc}$  defined in equation as

$$\beta_{ac} = \frac{\Delta I_C}{\Delta I_B} = \frac{i_c}{i_b}$$



which is also known as the ac current gain  $A_i$ . Usually  $\beta_{ac}$  is close to  $\beta_{dc}$  in the linear region of the output characteristics. The change in  $I_C$  due to a change in  $I_B$  causes a change in  $V_{CE}$  and the voltage drop across the resistor  $R_L$  because  $V_{CC}$  is fixed.

These changes can be given by Eq. as

$$\Delta V_{CC} = \Delta V_{CE} + R_L \Delta I_C = 0$$

or 
$$\Delta V_{CE} = -R_L \Delta I_C$$

The change in  $V_{CE}$  is the output voltage  $v_o$ . From equation we get

$$v_o = \Delta V_{CE} = -\beta_{ac} R_L \Delta I_B$$

The voltage gain of the amplifier is

$$A_v = \frac{v_o}{v_i} = \frac{\Delta V_{CE}}{r_i \Delta I_B} = -\frac{\beta_{ac} R_L}{r_i}$$

The negative sign represents that output voltage is opposite with phase with the input voltage.

From the discussion of the transistor characteristics you have seen that there is a current gain  $\beta_{ac}$  in the CE configuration. Here we have also seen the voltage gain  $A_v$ . Therefore the power gain  $A_p$  can be expressed as the product of the current gain and voltage gain. Mathematically

$$A_p = \beta_{ac} \times A_v$$

Since  $\beta_{ac}$  and  $A_v$  are greater than 1, we get ac power gain. However it should be realised that transistor is not a power generating device. The energy for the higher ac power at the output is supplied by the battery.

**Note: Transistor as CE amplifier**

(i) ac current gain  $\beta_{ac} = \left( \frac{\Delta I_C}{\Delta I_B} \right)_{V_{CE} = \text{constant}}$

(ii) dc current gain  $\beta_{dc} = \frac{i_c}{i_b}$

(iii) Voltage gain :  $A_v = \frac{\Delta V_o}{\Delta V_i} = \beta_{ac} \times \text{Resistance gain}$

(iv) Power gain =  $\frac{\Delta P_o}{\Delta P_i} = \beta_{ac}^2 \times \text{Resistance}$

(v) Transconductance ( $g_m$ ) : The ratio of the change in collector current to the change in emitter base voltage is

called trans conductance. i.e.  $g_m = \frac{\Delta i_c}{\Delta V_{EB}}$ . Also  $g_m = \frac{A_v}{R_L}$ ;  $R_L$  = Load resistance.

(3) Relation between  $\alpha$  and  $\beta$  :  $\beta = \frac{\alpha}{1-\alpha}$  or  $\alpha = \frac{\beta}{1+\beta}$

## PHYSICS FOR JEE MAIN & ADVANCED

**Ex.** Let  $i_E$ ,  $i_C$  and  $i_B$  represent the emitter current, the collector current and the base current respectively in a transistor. Then

(1)  $i_C$  is slightly smaller than  $i_E$ .

(2)  $i_C$  is slightly greater than  $i_E$ .

(3)  $i_B$  is much smaller than  $i_E$ .

(4)  $i_B$  is much greater than  $i_E$ .

**Ans.** (1,3)

**Ex.** In a common base transistor amplifier, the input and the output resistance are  $500\ \Omega$  and  $40\text{k}\Omega$ , and the emitter current is  $1.0\text{mA}$ . Find the input and the output voltages. Given  $\alpha = 0.95$ .

**Sol.** The input voltage is emitter current multiplied by input resistance, that is,

$$V_{in} = i_E \times R_{in} = (1.0 \times 10^{-3}\text{ A}) \times 500\Omega = 0.5\text{ V}$$

Similarly, the output voltage is

$$V_{out} = i_C \times R_{out} = \alpha i_E \times R_{out} \\ = 0.95 (1.0 \times 10^{-3}\text{ A}) \times (40 \times 10^3\ \Omega) = 38\text{ V}.$$

**Ex.** A P–N–P transistor is used in common–emitter mode in an amplifier circuit. A change of  $40\mu\text{A}$  in the base current brings a change of  $2\text{mA}$  in collector current and  $0.04\text{ V}$  in base–emitter voltage. Find the: (i) input resistance ( $R_{in}$ ), and (ii) the base current amplification factor ( $\beta$ ).

If a load of  $6\text{k}\Omega$  is used, then also find the voltage gain of the amplifier.

**Sol. :** Given  $\Delta I_B = 40\mu\text{A} = 40 \times 10^{-6}\text{ A}$

$$\Delta I_C = 2\text{mA} = 2 \times 10^{-3}\text{ A}$$

$$\Delta V_{BE} = 0.04\text{ volt}, R_L = 6\text{k}\Omega = 6 \times 10^3\ \Omega$$

(i) Input Resistance,  $R_{in} = \frac{\Delta V_{BE}}{\Delta I_B} = \frac{0.04}{40 \times 10^{-6}} = 10^3\ \Omega = 1\text{ k}\Omega$

(ii) Current amplification factor,  $\beta = \frac{\Delta I_C}{\Delta I_B} = \frac{2 \times 10^{-3}}{40 \times 10^{-6}} = 50$

(iii) Voltage gain in common–emitter configuration,

$$A_v = \beta \frac{R_L}{R_{in}} = 50 \times \frac{6 \times 10^3}{1 \times 10^3} = 300.$$

**Ex.** In an N–P–N transistor  $10^{10}$  electrons enter the emitter in  $10^{-6}\text{ s}$ . 2% of the electrons are lost in the base. Calculate the current transfer ratio and current amplification factor.

**Sol.** We know that current = charge/time

The emitter current ( $I_E$ ) is given by  $I_E = \frac{Ne}{t} = \frac{10^{10} \times (1.6 \times 10^{-19})}{10^{-6}} = 1.6\text{ mA}$

The base current ( $I_B$ ) is given by

$$I_B = \frac{2}{100} \times 1.6 = 0.032\text{ mA}$$

In a transistor,

$$I_E = I_B + I_C$$

$$I_C = I_E - I_B = 1.6 - 0.032 = 1.568\text{ mA}$$

Current transfer ratio =  $\frac{I_C}{I_E} = \frac{1.568}{1.6} = 0.98$

Current amplification factor =  $\frac{I_C}{I_B} = \frac{1.568}{0.032} = 49.$





**Ex.** When the voltage between emitter and the base  $V_{EB}$  of a transistor is changed by 5mV while keeping the collector voltage  $V_{CE}$  fixed when then its emitter current changes by 0.15 mA. Calculate the input resistance of the transistor.

**Ans.** 33.33 ohm

**Ex.** A transistor is used in common-emitter mode in an amplifier circuit. When a signal of 20 mV is added to the base-emitter voltage, the base current changes by  $20\mu\text{A}$  and the collector current changes by 2 mA. The load resistance is 5 k $\Omega$ . Calculate (a) the factor  $\beta$ , (b) the input resistance  $R_{BE}$ , (c) the transconductance and (d) the voltage gain.

**Sol.** (a)  $\beta = \frac{\Delta I_C}{\Delta I_B} = \frac{2\text{mA}}{20\mu\text{A}} = 100$

(b) The input resistance  $R_{BE} = \frac{\Delta V_{BE}}{\Delta I_B} = \frac{20\text{mV}}{20\mu\text{A}} = 1\text{k}\Omega$

(c) Transconductance  $= \frac{\Delta I_C}{\Delta V_{BE}} = \frac{2\text{mA}}{20\text{mV}} = 0.1 \text{ mho.}$

(d) The change in output voltage is  $R_L \Delta I_C$   
 $= (5 \text{ k}\Omega) (2\text{mA}) = 10\text{V}.$

The applied signal voltage = 20 mV.

Thus, the voltage gain is,

$$\frac{10\text{V}}{20\text{mV}} = 500.$$

**Ex.** The a-c current gain of a transistor is  $\beta = 19$ . In its common-emitter configuration, what will be the change in the collector-current for a change of 0.4 mA in the base-current ? What will be the change in the emitter current ?

**Sol.** By definition, the a-c current gain  $\beta$  is given by

$$\beta_{ac} = \frac{\Delta I_C}{\Delta I_B}$$

$$\therefore \Delta I_C = \beta \times \Delta I_B = 19 \times 0.4 \text{ mA} = 7.6 \text{ mA}.$$

The emitter - current is the sum of the base- current and the collector-current ( $i_E = i_B + i_C$ )

$$\therefore \Delta I_E = \Delta I_B + \Delta I_C = 0.4 \text{ mA} + 7.6 \text{ mA} = 80 \text{ mA}.$$

**Ex.** A transistor is connected in common-emitter (C-E) configuration. The collector-supply is 8 V and the voltage drop across a resistor of 800  $\Omega$  in the collector circuit is 0.5 V. If the current-gain factor ( $\alpha$ ) is 0.96, find the base-current.

**Sol.** The alternating-current gain is

$$\beta = \frac{\alpha}{1 - \alpha} = \frac{0.96}{1 - 0.96} = 24$$

The collector - current is

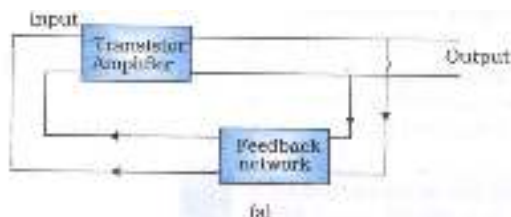
$$i_C = \frac{\text{voltage - drop across collector resistor}}{\text{resistance}} = \frac{0.5\text{V}}{800\Omega} \times 10^{-3} \text{ A}.$$

But  $\beta = \frac{i_C}{i_B}$ , where  $i_B$  is base - current.

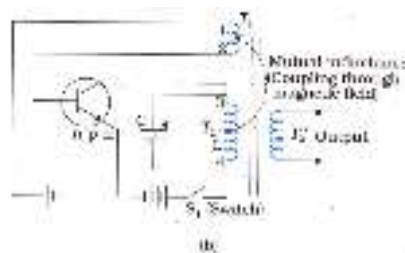
$$\therefore i_B = \frac{i_C}{\beta} = \frac{0.625 \times 10^{-3} \text{ A}}{24} = 26 \times 10^{-6} \text{ A} = 26 \mu\text{A}.$$

## 8. Feedback amplifier and transistor oscillator:

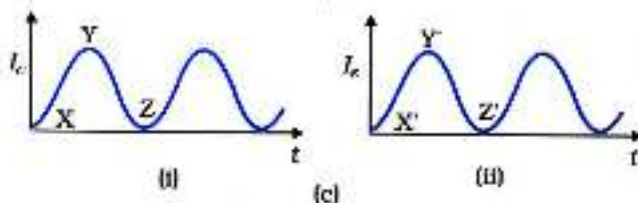
In an oscillator, we get ac output without any external input signal. A portion of the output power is returned back (feedback) to the input in phase with the starting power (this process is termed positive feedback) as shown in figure(a). The feedback can be achieved by inductive coupling (through mutual inductance) or LC or RC networks.



Suppose switch  $S_1$  is put on to apply proper bias for the first time. Obviously, a surge of collector current flows in the transistor. This current flows through the coil  $T_2$  where terminals are numbered 3 and 4 (Fig. b).



This current does not reach full amplitude instantaneously but increases from X To Y, as shown in figure(C). The inductive coupling between coil  $T_2$  and coil  $T_1$  now causes a current to flow in the emitter circuit (note that this actually is the 'feedback' from input to output). As a result of this positive feedback, this current (in  $T_1$  emitter current) also increases from X' to Y' Fig. (C) (ii).



The current in  $T_2$  (collector current), connected in the collector circuit acquires the value Y when the transistor becomes saturated. This means that maximum collector current is flowing and can increase no further. Since there is no further change in collector current, the magnetic field around  $T_2$  ceases to grow. As soon as the field becomes static, there will be no further feedback from  $T_2$  to  $T_1$ . Without continued feedback, the emitter current begins to fall. Consequently, collector current decreases causes the magnetic field to decay around the coil  $T_2$ . Thus,  $T_1$  is now seeing a decaying field in  $T_2$  (opposite from what it saw when the field was growing at the initial start operation). This causes a further decrease in means that both  $I_e$  and  $I_c$  cease to flow. Therefore, the transistor has reverted back to its original state (when the power was first switched on). The whole process now repeat itself. The transistor is driven to saturation, then to cut-off, and then back to saturation. The time for change from saturation to cut-off and back is determined by the constant of the tank circuit or tuned circuit (inductance L of Coil  $T_2$  and C connected in parallel to it). The resonance frequency ( $\nu$ ) of this tuned circuit determines the frequency at which the oscillator will oscillate.

$$\nu = \frac{1}{2\pi\sqrt{LC}}$$

## 9. Analogue Circuits and Digital Circuits and signal :

There are two types of electronic circuits : analogue circuits and digital circuits :

In analogue circuits, the voltage (or current) varies continuously with time (figure a). Such a voltage (or current) signal is called an 'analogue signal'. Figure shows a typical voltage analogue signal varying sinusoidally between 0 and 5V.

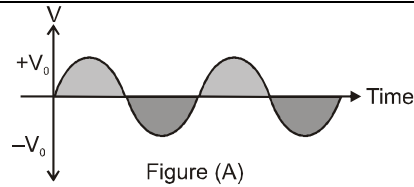


Figure (A)

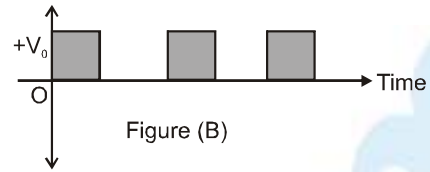


Figure (B)

On the other hand, in digital circuits, the voltage (or current) has only two levels, either zero or some constant value of voltage (figure b). A signal having only two levels of voltage (or current) is called a 'digital signal'. Figure shows a typical digital signal in which the voltage at any time is either 0 or 5V.

In digital circuits, the binary number system is used, according to which the two levels of the (digital) signal are represented by the digits 0 and 1 only.

The digital circuits are the basis of calculators, computers, etc.

## 10. Logic Gates :

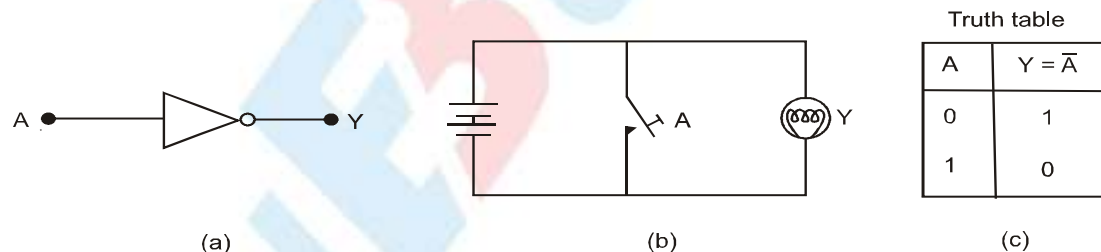
A logic gate is a digital circuit which works according to some logical relationship between input and output voltages. It either allows a signal to pass through or stops it.

A gate is a digital circuit that follows certain logical relationship between the input and output voltages. Therefore, they are generally known as logic gates — gates because they control the flow of information. The five common logic gates used are NOT, AND, OR, NAND, NOR. Each logic gate is indicated by a symbol and its function is defined by a truth table that shows all the possible input logic level combinations with their respective output logic levels. Truth tables help understand the behaviour of logic gates. These logic gates can be realised using semiconductor devices.

### (a) The NOT Gate :

The NOT gate has only one input and one output. It combines the input A with the output Y, according to the Boolean expression  $\bar{A} = Y$ ,

read as 'NOT A equals Y'. It means that Y is negation (or inversion) of A. Since there are only two digits 0 and 1 in the binary system, we have,  $Y = 0$ , if  $A = 1$  and  $Y = 1$  if  $A = 0$ . The logic symbol of the NOT gate is shown in figure.



The possible combinations of the input A and the output Y of the NOT gate can be known with the help of electric circuit, shown in figure. In this circuit, a switch A (input) is connected in parallel to a battery and a bulb Y (output). The working of the circuit is as follows :

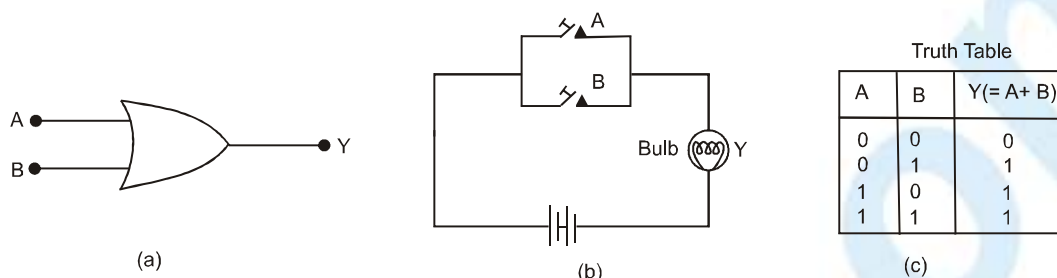
If switch A is open ( $A = 0$ ), the bulb will glow ( $Y = 1$ ).

If switch A is closed ( $A = 1$ ), the bulb will not glow ( $Y = 0$ ).

These two possible combinations of input A and output Y are tabulated in figure, which is the truth table of the NOT gate.

### (b) The OR Gate :

The OR gate is a device that has two input variables A and B and one output variable Y, and follows the Boolean expression,  $A + B = Y$ , read as 'A OR B equal Y'. Its logic symbol is shown in figure.



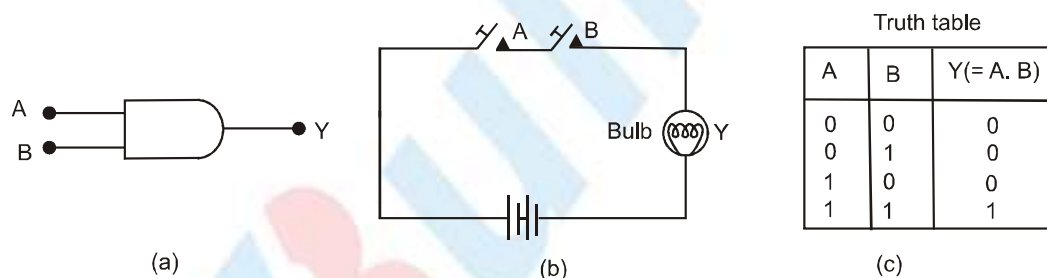
The possible combinations of the inputs A and B and the output Y of the OR gate can be known with the help of an electrical circuit, shown in figure. In this circuit, two switches A and B (inputs) are connected in parallel with a battery and a bulb Y (output).

### (c) The AND Gate :

The AND gate is also a two-input and one-output logic gate. It combines the inputs A and B to give the output Y, according to the Boolean expression

$$A \cdot B = Y$$

read 'A AND B equals Y'

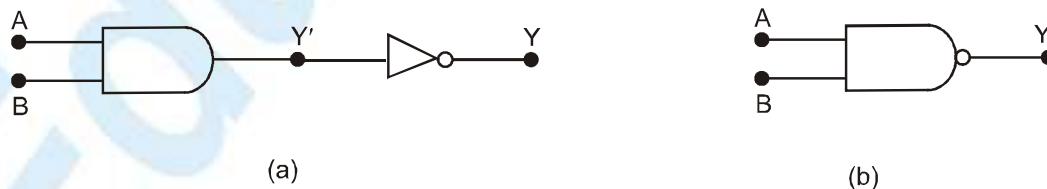


### Combinations of gates :

Various combinations of the three basic gates, namely, OR, AND and NOT, produce complicated digital circuits, which are also called 'gates'. The commonly used combinations of basic gates are NAND gate, NOR, gate. These are also called universal gates.

#### (i) The NAND gate :

This gate is a combination of AND and NOT gates. If the output Y' of AND gate is connected to the input of NOT gate, as shown in figure, the gate so obtained is called NAND gate. The logic symbol of NAND gate is shown in figure.



The Boolean expression for the NAND gate is  $\overline{A \cdot B} = Y$  read as 'A AND B negated equals Y'.

The truth table of the NAND gate can be obtained by logically combining the truth tables of AND and NOT gates. In figure, the output  $Y'$  of the truth table of AND gate have been negated (NOT operation) to obtain the corresponding outputs  $Y$  for the NAND gate. The resulting table is the truth table of the NAND gate.

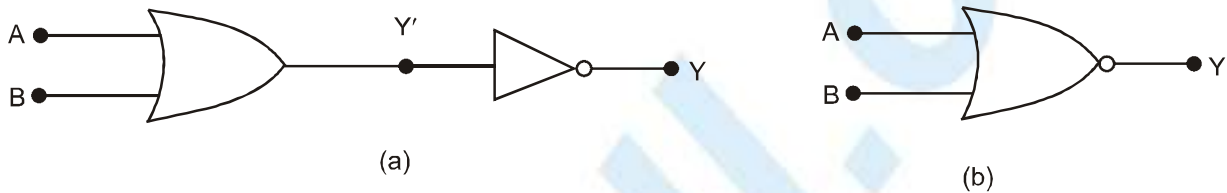
A	B	$Y' (= A \cdot B)$	$Y (= \overline{A \cdot B}) = \overline{Y'}$
0	0	0	1
0	1	0	1
1	0	0	1
1	1	1	0

 $\Rightarrow$ 

A	B	Y
0	0	1
0	1	1
1	0	1
1	1	0

### (ii) The NOR Gate :

The NOR gate is a combination of OR and NOT gates. If the output  $Y'$  of OR gate is connected to the input of NOT gate, the gate so obtained is NOR gate.



The Boolean expression for the NOR gate is

$$\overline{A + B} = Y$$

read as 'A OR B negated equals Y' :

A	B	$Y' (= A + B)$	$Y (= \overline{A + B}) = \overline{Y'}$
0	0	0	1
0	1	1	0
1	0	1	0
1	1	1	0

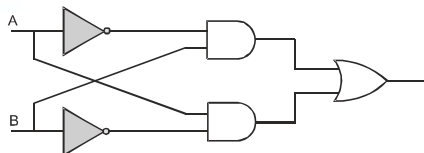
A	B	Y
0	0	1
0	1	0
1	0	0
1	1	0

The truth table of the NOR gate can be obtained by logically combining the truth tables of OR and NOT gates. In figure(a), the outputs  $Y'$  of the truth table of OR gate have been negated to obtain the corresponding outputs  $Y$  for the NOR gate.

### (iii) The XOR Gate :

The Boolean expression for the XOR gate is

$$Y = A \cdot \overline{A \cdot B} + \overline{A \cdot B} \cdot B$$

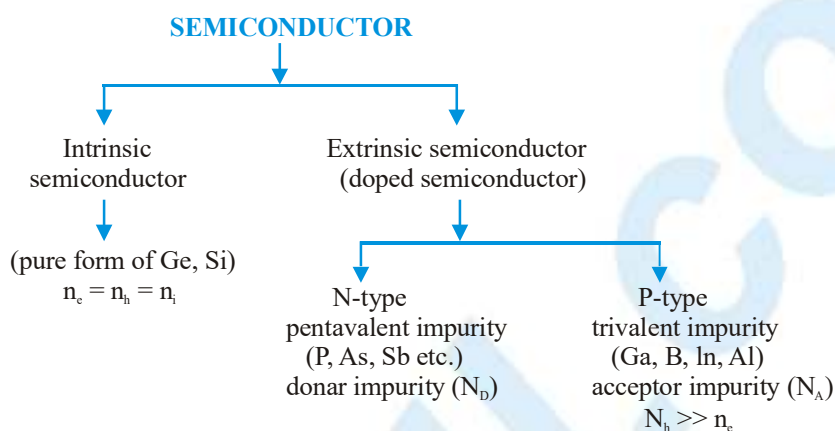




1. Number of electrons reaching from valence band to conduction band

$$n = AT^{3/2} e^{-\frac{\Delta E_g}{2kT}}$$

2. Classification of Semiconductors :



3. Mass-action law

$$n_i^2 = n_e \times n_h$$

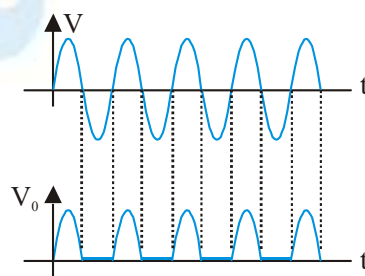
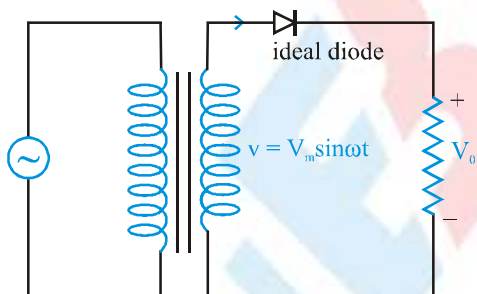
(a) For N-type semiconductor  $n_e = N_D$

(b) For P-type semiconductor  $n_h = N_A$

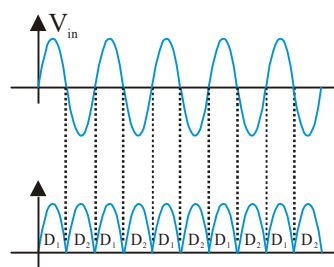
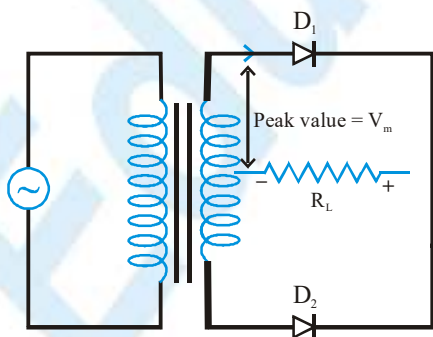
4. Conductivity

$$n_i e (\mu_e + \mu_h)$$

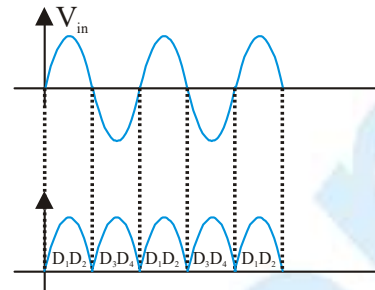
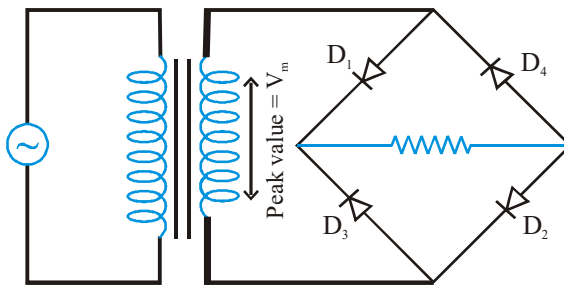
5. Half wave rectifier



6. Centre – Tap Full wave Rectifier



7. Full wave Bridge rectifier



8. Form factor =  $\frac{I_{rms}}{I_{dc}}$

(a) For HWR (Half wave rectifier) Form factor =  $\frac{\pi}{2}$

(b) For FWR (Full wave rectifier) Form factor =  $\frac{\pi}{2\sqrt{2}}$

9. Ripple factor  $r = \frac{I_{ac}}{I_{dc}}$

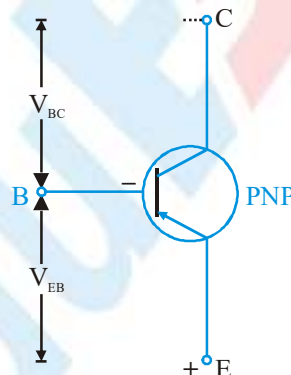
(a) For HWR  $r = 1.21$  (b) For FWR  $r = 0.48$

10. Rectifier efficiency  $\eta = \frac{P_{dc}}{P_{ac}} = \frac{I_{dc}^2 R_L}{I_{rms}^2 (R_F + R_L)}$

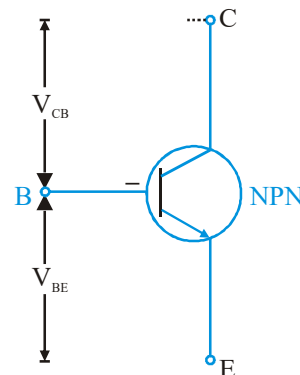
(a) For HWR  $\eta\% = 1 + \frac{R_F}{R_L} \times 40.6$  & FWR  $\eta\% = \frac{81.2}{1 + \frac{R_F}{R_L}}$

11. For transistor

$$I_E = I_B + I_C$$



(a)



(b)

Comparative study of transistor configurations

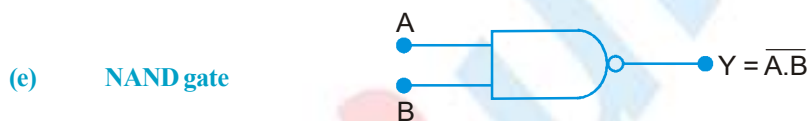
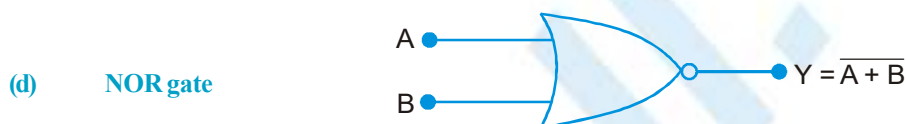
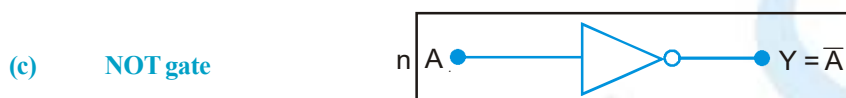
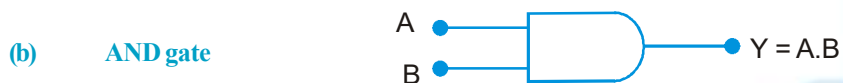
1. Common Base (CB)

3. Common Collector (CC)

2. Common Emitter (CE)

12. Relation between  $\alpha$ ,  $\beta$  and  $\gamma$ :  $\beta = \frac{\alpha}{1-\alpha}$ ,  $\gamma = 1 + \beta$ ,  $\gamma = \frac{1}{1-\alpha}$

13. Logic gates



14. De Morgan's theorem  $\overline{A+B} = \bar{A} \cdot \bar{B}$ ,  $\overline{A \cdot B} = \bar{A} + \bar{B}$